# Lecture 5

## Implementation of Linear Timeinvariant Systems

## **Different Implementation OPtions**

- We will mainly discuss the following structures:
- Direct Form Implementation,
- Cascade Form Implementation (Cascade Structure),
- Parallel Structure,
- Lattice Structure,

## **Direct Implementation**

Consider the general form of a linear time invariant system (filter). That is one with both feedback and feed forward path. The linear difference equation describing such a filter is:

$$y(n) = -\sum_{k=1}^{N} a_k y(n-k) + \sum_{k=0}^{M} b_k x(n-k)$$

Using z-transform, we find the transfer function of the system as:

$$H(z) = \frac{\sum_{k=0}^{M} b_k z^{-k}}{1 + \sum_{k=1}^{N} a_k z^{-k}}$$

## Structure of FIR Systems

Assume that there is no feedback, i.e., the output y[n] does not depend on the past outputs and only depends on the input sequence:

$$y(n) = \sum_{k=0}^{M-1} b_k x(n-k)$$

> The denominator of H(z) will be equal to one and we will have,

$$H(z) = \sum_{k=0}^{M-1} b_k z^{-k}$$

▶ The unit sample response of the FIR system will be:

$$h(n) = \begin{cases} b_n, & 0 \le n \le M - 1\\ 0, & \text{otherwise} \end{cases}$$

## **Direct Implementation of FIR Systems**



When the FIR system has linear phase (it is symmetric):

$$h(n) = \pm h(M - 1 - n)$$

## Linear Phase FIR Systems

When the FIR system has linear phase (it is symmetric):

$$h(n) = \pm h(M - 1 - n)$$



## **Exercise:**

- 1) Implement y[n] = 3x[n] 2x[n-1] + x[n-2].
- 2) Is the filter y[n]=x[n]-2x[n-1]-x[n-2] linear phase?

3) Implement the filter:

y[n] = 0.5x[n] + x[n-1] + 2x[n-2] + x[n-3] + 0.5x[n-4]

## **Cascade Implementation of FIR Systems**

It is easy to factor H(z) into K second order filters:

$$H(z) = \prod_{k=1}^{K} H_k(z)$$

▶ where,

$$H_k(z) = b_{k0} + b_{k1}z^{-1} + b_{k2}z^{-2}, \qquad k = 1, 2, \dots, K$$

▶ and K is the M/2 for M even and (M+1)/2 for M odd.



## **Cascade Implementation of FIR Systems**

▶ Where each section,

$$H_k(z) = b_{k0} + b_{k1}z^{-1} + b_{k2}z^{-2}, \qquad k = 1, 2, \dots, K$$

Is implemented as



Lattice filter implementation is widely used in adaptive filtering. Assume that we have a filter with transfer function H(z). We can write,

$$H_m(z) = A_m(z), \qquad m = 0, 1, 2, \dots, M-1$$

• where  $A_m(z)$  is a polynomial with  $A_0(z) = 1$ 

$$A_m(z) = 1 + \sum_{k=1}^m \alpha_m(k) z^{-k}, \qquad m \ge 1$$

> Y[n] can be written as

$$y(n) = x(n) + \sum_{k=1}^{m} \alpha_m(k) x(n-k)$$

> The direct form implementation can be expressed as



x(n)  $z^{-1}$   $z^{$ 

This is called a prediction error filter.

- Let's consider a first order FIR filter, i.e., m=1:  $y(n) = x(n) + \alpha_1(1)x(n-1)$
- Let the reflection coefficient  $K_1 = \alpha_1(1)$ . to get:



Now consider m=2:

 $y(n) = x(n) + \alpha_2(1)x(n-1) + \alpha_2(2)x(n-2)$ 

- We cascade two lattice stages:
- The out put of the first stage is,

$$f_1(n) = x(n) + K_1 x(n-1)$$

$$g_1(n) = K_1 x(n) + x(n-1)$$

And the output of the second stage is:

$$f_2(n) = f_1(n) + K_2 g_1(n-1)$$
$$g_2(n) = K_2 f_1(n) + g_1(n-1)$$



Two-stage lattice filter.

• Let's consider on  $f_2[n]$ :

$$f_2(n) = x(n) + K_1 x(n-1) + K_2 [K_1 x(n-1) + x(n-2)]$$

$$= x(n) + K_1(1 + K_2)x(n-1) + K_2x(n-2)$$

•  $f_2[n]$  will be y[n] if:

$$\alpha_2(2) = K_2, \qquad \alpha_2(1) = K_1(1+K_2)$$

or, equivalently if:

$$K_2 = \alpha_2(2), \qquad K_1 = \frac{\alpha_2(1)}{1 + \alpha_2(2)}$$



### Conversion of reflection coefficients to filter taps

Use the following equations recursively,

$$A_0(z) = B_0(z) = 1$$
  

$$A_m(z) = A_{m-1}(z) + K_m z^{-1} B_{m-1}(z), \qquad m = 1, 2, \dots, M - 1$$
  

$$B_m(z) = z^{-m} A_m(z^{-1}), \qquad m = 1, 2, \dots, M - 1$$

#### EXAMPLE

Given a three-stage lattice filter with coefficients  $K_1 = \frac{1}{4}$ ,  $K_2 = \frac{1}{4}$ ,  $K_3 = \frac{1}{3}$ , determine the FIR filter coefficients for the direct-form structure.

Solution. 
$$m = 1$$
. Thus we have  $A_1(z) = A_0(z) + K_1 z^{-1} B_0(z)$   
=  $1 + K_1 z^{-1} = 1 + \frac{1}{4} z^{-1}$ 

Hence the coefficients of an FIR filter corresponding to the single-stage lattice are  $\alpha_1(0) = 1$ ,  $\alpha_1(1) = K_1 = \frac{1}{4}$ . Since  $B_m(z)$  is the reverse polynomial of  $A_m(z)$ , we have

$$B_1(z) = \frac{1}{4} + z^{-1}$$

### Conversion of reflection coefficients to filter taps

Next we add the second stage to the lattice. For m = 2,

$$A_2(z) = A_1(z) + K_2 z^{-1} B_1(z)$$
$$= 1 + \frac{3}{8} z^{-1} + \frac{1}{2} z^{-2}$$

Hence the FIR filter parameters corresponding to the two-stage lattice are  $\alpha_2(0) = 1$ ,  $\alpha_2(1) = \frac{3}{8}$ ,  $\alpha_2(2) = \frac{1}{2}$ . Also,  $B_2(z) = \frac{1}{2} + \frac{3}{8}z^{-1} + z^{-2}$ 

Finally, the addition of the third stage to the lattice results in the polynomial

$$A_3(z) = A_2(z) + K_3 z^{-1} B_2(z)$$
$$= 1 + \frac{13}{24} z^{-1} + \frac{5}{8} z^{-2} + \frac{1}{3} z^{-1}$$

Consequently, the desired direct-form FIR filter is characterized by the coefficients

$$\alpha_3(0) = 1, \qquad \alpha_3(1) = \frac{13}{24}, \qquad \alpha_3(2) = \frac{5}{8}, \qquad \alpha_3(3) = \frac{1}{3}$$

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### Conversion of FIR Taps to Lattice reflection coefficients

• We start with  $A_{M-1}(z)$  and find lattice coefficients using

$$A_m(z) = A_{m-1}(z) + K_m z^{-1} B_{m-1}(z)$$
  
=  $A_{m-1}(z) + K_m [B_m(z) - K_m A_{m-1}(z)]$ 

Or, equivalently,

$$A_{m-1}(z) = \frac{A_m(z) - K_m B_m(z)}{1 - K_m^2}, \qquad m = M - 1, M - 2, \dots, 1$$

#### EXAMPLE

Determine the lattice coefficients corresponding to the FIR filter with system function

$$H(z) = A_3(z) = 1 + \frac{13}{24}z^{-1} + \frac{5}{8}z^{-2} + \frac{1}{3}z^{-3}$$

**Solution.** First we note that  $K_3 = \alpha_3(3) = \frac{1}{3}$ . Furthermore,

$$B_3(z) = \frac{1}{3} + \frac{5}{8}z^{-1} + \frac{13}{24}z^{-2} + z^{-3}$$

### Conversion of FIR Taps to Lattice reflection coefficients

The step-down relationship with m = 3 yields

$$A_2(z) = \frac{A_3(z) - K_3 B_3(z)}{1 - K_3^2}$$
$$= 1 + \frac{3}{8} z^{-1} + \frac{1}{2} z^{-2}$$

Hence  $K_2 = \alpha_2(2) = \frac{1}{2}$  and  $B_2(z) = \frac{1}{2} + \frac{3}{8}z^{-1} + z^{-2}$ . By repeating the step-down recursion we obtain

Hence 
$$K_1 = \alpha_1(1) = \frac{1}{4}$$
.  
 $A_1(z) = \frac{A_2(z) - K_2 B_2(z)}{1 - K_2^2}$   
 $= 1 + \frac{1}{4} z^{-1}$ 

A recursive formula for findingthe lattice coefficients:

$$\alpha_{m-1}(k) = \frac{\alpha_m(k) - K_m \beta_m(k)}{1 - K_m^2}$$

$$=\frac{\alpha_m(k)-\alpha_m(m)\alpha_m(m-k)}{1-\alpha_m^2(m)}, \qquad 1\le k\le m-1$$

### Structure of IIR Systems

An IIR System can be expressed as  $H(z) = H_1(z)H_2(z)$ 

With an all-zero part

$$H_1(z) = \sum_{k=0}^M b_k z^{-k}$$

and an all-pole part

$$H_2(z) = \frac{1}{1 + \sum_{k=1}^{N} a_k z^{-k}}$$

### Structure of IIR Systems: Direct form I



### Structure of IIR Systems: Direct form II

- If we put the all-pole filter firs, we get
- w[n] is the output of  $H_2(z)$  and
- the input to  $H_1(z)$
- ► So, we have

$$w(n) = -\sum_{k=1}^{N} a_k w(n-k) + x(n)$$

and

$$y(n) = \sum_{k=0}^{M} b_k w(n-k)$$



### Structure of IIR Systems: Cascade Implementation

► The transfer function H(z) can be written as:

$$H(z) = \prod_{k=1}^{K} H_k(z)$$

- ▶ where K is [N/2]. That is it is N/2 for N even and (N+1)/2 for N odd.
- $H_k(z)$  has the general form of:

$$H_k(z) = \frac{b_{k0} + b_{k1}z^{-1} + b_{k2}z^{-2}}{1 + a_{k1}z^{-1} + a_{k2}z^{-2}}$$

### Structure of IIR Systems: Cascade Implementation



Cascade structure of second-order systems and a realization of each second-order section.

## Structure of IIR Systems: Parallel Implementation

Performing partial fraction expansion, we get,  $H(z) = C + \sum_{k=1}^{N} \frac{A_k}{1 - p_k z^{-1}}$ 

so we can implement the filter using N parallel branches each with a single pole.

Most often the poles are complex. The complex poles are pairs of conjugate poles. So, we can implement the filter with parallel second order branches:



### Structure of IIR Systems: Parallel Implementation



## Example:

Find the cascade and parallel implementation of the filter:

$$H(z) = \frac{10(1 - \frac{1}{2}z^{-1})(1 - \frac{2}{3}z^{-1})(1 + 2z^{-1})}{(1 - \frac{3}{4}z^{-1})(1 - \frac{1}{8}z^{-1})[1 - (\frac{1}{2} + j\frac{1}{2})z^{-1}][1 - (\frac{1}{2} - j\frac{1}{2})z^{-1}]}$$

We have

$$H_2(z) = \frac{1 + \frac{3}{2}z^{-1} - z^{-2}}{1 - z^{-1} + \frac{1}{2}z^{-2}}$$

 $H_1(z) = \frac{1 - \frac{2}{3}z^{-1}}{1 - \frac{7}{2}z^{-1} + \frac{3}{22}z^{-2}}$ 

So

 $H(z) = 10H_1(z)H_2(z)$ 

## Example:

And the cascade implementation is:



## Example:

For parallel implementation, we have:

$$H(z) = \frac{A_1}{1 - \frac{3}{4}z^{-1}} + \frac{A_2}{1 - \frac{1}{8}z^{-1}} + \frac{A_3}{1 - \frac{1}{8}z^{-1}} + \frac{A_3^*}{1 - (\frac{1}{2} + j\frac{1}{2})z^{-1}} + \frac{A_3^*}{1 - (\frac{1}{2} - j\frac{1}{2})z^{-1}}$$



### Structure of IIR Systems: Lattice Implementation

We have already done the all-zero filter:



### Structure of IIR Systems: Lattice Implementation

Now, let's consider the all-pole filter:

$$H_2(z) = \frac{1}{1 + \sum_{k=1}^{N} a_k z^{-k}}$$
$$H_2(z) = \frac{1}{1 + \sum_{k=1}^{N} a_N(k) z^{-k}} = \frac{1}{A_N(z)}$$

We can write y[n] as:

or,

$$y(n) = -\sum_{k=1}^{N} a_N(k)y(n-k) + x(n)$$

Structure of IIR Systems: Lattice Implementation Then x[n] can be written as:  $x(n) = y(n) + \sum_{k=1}^{N} a_N(k)y(n-k)$ 

 $f_N(n) = x(n)$ 

This is a lattice filter with similar to the one we saw before but the role of input and output changed. That is we have  $x(n) = f_N(n)$  and  $y(n) = f_0(n)$  We use:

$$f_{m-1}(n) = f_m(n) - K_m g_{m-1}(n-1), \qquad m = N, N-1, \dots, 1$$
$$g_m(n) = K_m f_{m-1}(n) + g_{m-1}(n-1), \qquad m = N, N-1, \dots, 1$$
$$v(n) = f_0(n) = g_0(n)$$

Input



Lattice structure for an all-pole IIR system.

### Lattice Ladder Implementation of IIR Systems



Lattice-ladder structure for the realization of a pole-zero system.