## ELEC 6131: Error Detecting and Correcting Codes

## Instructor:

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## LECTURE 6: BCH Codes

## BCH Codes

Block Length $\mathrm{n}=2^{\mathrm{m}}-1$ for some $\mathrm{m} \geq 3$
Number of Parity-check bits $n-k \leq m t$ Minimum Distance $\mathrm{d}_{\text {min }} \geq 2 t+1$

- The generator polynomial is defined in terms of its roots over GF $\left(2^{\mathrm{m}}\right)$.
- For a t -error correcting BCH Code, $\mathrm{g}(\mathrm{x})$ is the lowest-degree polynomial with $\operatorname{roots} \alpha, \alpha^{2} \ldots, \alpha^{2 t}$.
$\Rightarrow$ Let $\varphi_{i}(x)$ be the minimal polynomial of $\alpha^{i}$ for $i=1,2, \ldots, 2 t$.Then:

$$
g(x)=\operatorname{LCM}\left\{\varphi_{1}(x), \varphi_{2}(x), \ldots, \varphi_{2 t}(x)\right\}
$$

Where LCM stands for least Common Multiple.

## BCH Codes

If $i$ is even then we can write $i=i^{\prime} .2^{l}$,
Where $i^{\prime}$ is odd and $l \geq 1$. Then:

$$
\alpha^{i}=\left(\alpha^{i^{\prime}}\right)^{2 l}
$$

So $\alpha^{i}$ and $\alpha^{i^{\prime}}$ are conjugate of each other and have the same minimal polynomial and:

$$
g(x)=\operatorname{LCM}\left\{\varphi_{1}(x), \varphi_{3}(x), \ldots, \varphi_{2 t-1}(x)\right\}
$$

- Since the degree of each of $\Phi_{i}(x), i=1,3, \ldots$ is less than or equal to m , the degree $\mathrm{of} \mathrm{g}(\mathrm{x})$ is less than or equal to $m t$ So,

$$
n-k \leq m t
$$

as the degree of $\mathrm{g}(\mathrm{x})$ is $n-k$.

- Table 6.1 lists BCH Codes for lengths $2^{m}-1, m=3, . .10$ that is length 7 to 1023.
- Refer to Appendix C for the list of BCH Codes and their generating polynomial.
- These are narrow sense or primitive BCH Codes. In general, $\alpha$ does not need to be primitive and roots can be non- Consecutive.

TABLE 6.1: BCH codes generated by primitive elements of order less than $2^{10}$

| $\boldsymbol{n}$ | $\boldsymbol{k}$ | $\boldsymbol{t}$ | $\boldsymbol{n}$ | $\boldsymbol{k}$ | $\boldsymbol{t}$ | $\boldsymbol{n}$ | $\boldsymbol{k}$ | $\boldsymbol{t}$ |
| ---: | ---: | ---: | ---: | ---: | ---: | ---: | ---: | ---: |
| 7 | 4 | 1 | 127 | 50 | 13 | 255 | 71 | 29 |
| 15 | 11 | 1 |  | 43 | 14 |  | 63 | 30 |
|  | 7 | 2 |  | 36 | 14 |  | 55 | 31 |
|  | 5 | 3 |  | 29 | 21 |  | 47 | 42 |
| 31 | 26 | 1 |  | 22 | 23 |  | 45 | 43 |
|  | 21 | 2 |  | 15 | 27 |  | 37 | 45 |
|  | 16 | 3 |  | 8 | 31 |  | 29 | 47 |
|  | 11 | 5 | 255 | 247 | 1 |  | 21 | 55 |
| 6 | 7 |  | 239 | 2 |  | 13 | 59 |  |
| 63 | 57 | 1 |  | 231 | 3 |  | 9 | 63 |
|  | 51 | 2 |  | 223 | 4 | 511 | 502 | 1 |
|  | 45 | 3 |  | 215 | 5 |  | 493 | 2 |
|  | 39 | 4 |  | 207 | 6 |  | 484 | 3 |
|  | 36 | 5 |  | 199 | 7 |  | 475 | 4 |
| 30 | 6 |  | 191 | 8 |  | 466 | 5 |  |
|  | 24 | 7 |  | 187 | 9 |  | 457 | 6 |

TABLE 61: (contirnted)


TABLE 6.1: (continued)

| $\boldsymbol{n}$ | $\boldsymbol{k}$ | $\boldsymbol{t}$ | $\boldsymbol{n}$ | $\boldsymbol{k}$ | $\boldsymbol{t}$ | $\boldsymbol{n}$ |
| :---: | :---: | :---: | ---: | ---: | ---: | ---: |
| 728 | 30 | $\boldsymbol{k}$ | $\boldsymbol{t}$ |  |  |  |
| 718 | 31 | 433 | 74 | 153 | 125 |  |
| 708 | 34 | 423 | 75 | 143 | 126 |  |
| 698 | 35 | 413 | 77 | 133 | 127 |  |
| 688 | 36 |  | 393 | 78 | 123 | 170 |
| 678 | 37 |  | 383 | 82 | 121 | 171 |
| 668 | 38 |  | 378 | 83 | 111 | 173 |
| 658 | 39 |  | 368 | 85 | 91 | 175 |
| 648 | 41 |  | 358 | 86 | 86 | 181 |
| 638 | 42 | 348 | 87 | 76 | 187 |  |
| 628 | 43 | 338 | 89 | 66 | 189 |  |
| 618 | 44 | 328 | 90 | 56 | 191 |  |
| 608 | 45 | 318 | 91 | 46 | 219 |  |
| 598 | 46 | 308 | 93 | 36 | 223 |  |
| 588 | 47 | 298 | 94 | 26 | 239 |  |
| 578 | 49 | 288 | 95 | 16 | 147 |  |
| 573 | 50 | 278 | 102 | 11 | 255 |  |
| 563 | 51 |  |  |  |  |  |
|  |  |  |  |  |  |  |

## Relationship with Hamming Codes

- Consider a single error correcting BCH Code of length $\mathrm{n}=2^{\mathrm{m}}-1$. Then:

$$
g(x)=\varphi_{1}(x)
$$

- $\varphi_{1}(\mathrm{x})$ is polynomial of degree $m$. So,

$$
n-k=m \rightarrow k=2^{m}-1-m
$$

So, a Hamming Code is just a single error correcting BCH code.

## BCH Codes: Example

- Example: Design a triple error correcting BCH Code of length 15.

$$
n=15=2^{m}-1 \rightarrow m=4
$$

- So, we need to find primitive element $\alpha$ over $G F\left(2^{4}\right)$ and form:

$$
g(x)=\operatorname{LCM}\left\{\varphi_{1}(x), \varphi_{3}(x), \varphi_{5}(x)\right\}
$$

- From table 2.9, we have:

$$
\begin{gathered}
\varphi_{1}(x)=1+x+x^{4} \\
\varphi_{3}(x)=1+x+x^{2}+x^{3}+x^{4} \\
\varphi_{5}(x)=1+x+x^{2}
\end{gathered}
$$

So,

$$
\begin{gathered}
g(x)=\left(1+x+x^{4}\right)\left(1+x+x^{2}+x^{3}+x^{4}\right)\left(1+x+x^{2}\right) \\
=1+x+x^{2}+x^{4}+x^{5}+x^{8}+x^{10}
\end{gathered}
$$

Therefore, $n-k=10 \rightarrow(15,5)$ BCH Code with $d_{\text {min }}=7 \rightarrow \mathrm{t}=3$.

- See Appendix B for minimal polynomials for $m=2, \ldots, 10$.

TABLE 2.9: Minimal polynomials of the clements in $G F\left(2^{4}\right)$ generated by $p(X)=$ $x^{4}+x+1$.


## BCH Codes Over $G F\left(2^{6}\right)$

- Do this derivation of $\mathrm{g}(\mathrm{x})$ for all BCH Codes of length $2^{6}-1=63$ in order to become familiar with concepts involved.
- First, using the primitive polynomial $\mathrm{p}(\mathrm{x})=1+\mathrm{x}+\mathrm{x}^{6}$, generate all elements of $G F\left(2^{6}\right)$. They are listed below, but I strongly encourage you to create the table yourself manually (don't use a computer program).



## BCH Codes Over GF ( $\mathbf{2}^{6}$ )



## BCH Codes Over GF ( $2^{6}$ )

- From the above table you can find minimal polynomial for all elements of $\operatorname{GF}\left(2^{6}\right)$ :

| Elements | Minimal polynomials |
| :---: | :---: |
| $\alpha_{0}, \alpha^{2}, \alpha^{4}, \alpha^{16}, \alpha^{32}$ | $1+X+X^{6}$ |
| $\alpha^{3}, \alpha^{6}, \alpha^{12} \alpha^{24}, \alpha^{48} \alpha^{33}$ | $1+x+x^{2}+x^{4}+x^{6}$ |
| $\alpha^{5} \cdot \alpha^{10}, \alpha^{20}, \alpha^{40} \cdot \alpha^{17}, \alpha^{34}$ | $1+x+X^{2}+x^{5}+X^{6}$ |
| ${ }^{14} \cdot \alpha^{28}, \alpha^{56}, \alpha^{49} \cdot \alpha^{35}$ | $1+x^{3}+x^{6}$ |
| $\alpha^{9} \cdot \alpha^{18}{ }^{18} \cdot \alpha^{36}{ }^{36}{ }^{\text {a }}$ | $1+x^{2}+x^{3}$ |
| $\begin{aligned} & 111, \alpha^{22}, \alpha^{44}, \alpha^{25}, \alpha^{50} \cdot \alpha^{37} \\ & 13-\alpha^{26}, \alpha^{52}, \alpha^{41}, \alpha^{19} \cdot \alpha^{38} \end{aligned}$ | $1+x^{2}+x^{3}+x^{5}+x^{6}$ $1+x+x^{3}+x^{4}+x^{6}$ |
| $\alpha^{15}, \alpha^{30}, \alpha^{60}, \alpha^{57}, \alpha^{51}, \alpha^{39}$ | $1+x^{2}+x^{3}+x^{4}+x^{6}$ $1+x^{2}+x^{4}+x^{5}+x^{6}$ |
| $\alpha^{21}, \alpha^{42}$ | $1+x+x^{2}+x^{5}+x^{5}$ |
| $\alpha^{23} \cdot \alpha^{46} \cdot \alpha^{29} \cdot \alpha^{58} \cdot \alpha^{53}, \alpha^{43}$ | $1+x+x^{4}+x^{5}+x^{6}$ $1+x+x^{3}$ |
| $\alpha^{31}, \alpha^{62}, \alpha^{61}, \alpha^{59}, \alpha^{55}, \alpha^{47}$ | 1+ ${ }^{1+x+x^{5}+x^{6}}$ |

Finally for any value of $t$ generate

$$
g(x)=\operatorname{LCM}\left\{\varphi_{1}(x), \varphi_{3}(x), \ldots, \varphi_{2 t-1}(x)\right\}
$$

TABLE 6.4: Generator polynomials of all the BCH codes of length 63.

| $n$ | $k$ | $t$ | $g(X)$ |
| :---: | :---: | :---: | :---: |
| 63 | 57 | 1 | $\mathrm{g}_{1}(X)=1+X+X^{6}$ |
|  | 51 | 2 | $\mathrm{g}_{2}(X)=\left(1+X+X^{6}\right)\left(1+X+X^{2}+X^{4}+X^{6}\right)$ |
|  | 45 | 3 | $g_{3}(X)=\left(1+X+X^{2}+X^{5}+X^{6}\right)_{2}(X)$ |
|  | 39 | 4 | $g_{4}(X)=\left(1+X^{3}+X^{6}\right)^{\prime} \mathrm{g}_{3}(X)$ |
|  | 36 | 5 | $g_{s}(X)=\left(1+X^{2}+X^{3}\right)_{4}(X)$ |
|  | 30 | 6 | $\mathrm{g}_{6}(X)=\left(1+X^{2}+X^{3}+X^{5}+X^{6}\right) \operatorname{gs}(X)$ |
|  | 24 | 7 | $g_{7}(X)=\left(1+X+X^{3}+X^{4}+X^{6}\right) \mathrm{g}_{6}(X)$ |
|  | 18 | 10 | $\mathrm{g}_{10}(X)=\left(1+X^{2}+X^{4}+X^{5}+X^{6}\right) \mathrm{g}_{7}(X)$ |
|  | 16 | 11 | $g_{11}(X)=\left(1+X+X^{2}\right) \mathrm{g}_{10}(X)$ |
|  | 10 | 13 | $g_{13}(X)=\left(1+X+X^{4}+X^{5}+X^{6}\right)_{11}(X)$ |
|  | 7 | 15 | $\operatorname{gis}_{15}(X)=\left(1+X+X^{3}\right) \ln _{13}(X)$ |

## Parity Check Matrix of BCH Codes

- We know that each code polynomial $\mathrm{v}(\mathrm{x})$ is divisible by $\mathrm{g}(\mathrm{x})$ and that $\mathrm{g}(\mathrm{x})$ is:

$$
g(x)=\operatorname{LCM}\left\{g_{1}(x), g_{2}(x), \ldots, g_{2 t}(x)\right\}
$$

$>$ So, $\alpha, \alpha^{2}, \alpha^{3}, \ldots, \alpha^{2 t}$ are the root of $\mathrm{v}(\mathrm{x})$, i.e.,

$$
V\left(\alpha^{i}\right)=v_{0}+v_{1} \alpha^{i}+v_{2} \alpha^{2 i}+\ldots+v_{n-1} \alpha^{(n-1) i}=0
$$

for $i=1,2, \ldots, 2 t$

- If we form

$$
H=\left[\begin{array}{ccccc}
1 & \alpha & \alpha^{2} & \cdots & \alpha^{n-1} \\
1 & \alpha^{2} & \left(\alpha^{2}\right)^{2} & \cdots & \left(\alpha^{2}\right)^{n-1} \\
\vdots & \vdots & \vdots & & \cdots \\
\vdots \\
1 & \alpha^{2 t} & \left(\alpha^{2 t}\right)^{2} & \cdots & \left(\alpha^{2 t}\right)^{n-1}
\end{array}\right]
$$

we have

$$
\underline{v} \cdot H^{T}=\underline{0}
$$

for any code vector $\underline{v}=\left(v_{0}, v_{1}, \ldots, v_{n-1}\right)$

## Parity Check Matrix of BCH Codes

- Since if $\alpha^{i}$ is conjugate of $\alpha^{j}$ then $v\left(\alpha^{i}\right)=0$ implies $v\left(\alpha^{j}\right)=0$ and vice versa. So, we can drop even rows and write:

$$
\mathrm{H}=\left[\begin{array}{cccccc}
1 & \alpha & \alpha^{2} & \alpha^{3} & \cdots & \alpha^{n-1} \\
1 & \alpha^{3} & \left(\alpha^{3}\right)^{2} & \left(\alpha^{3}\right)^{3} & \cdots & \left(\alpha^{3}\right)^{n-1} \\
1 & \alpha^{5} & \left(\alpha^{5}\right)^{2} & \left(\alpha^{5}\right)^{3} & \cdots & \left(\alpha^{5}\right)^{n-1} \\
\vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\
1 \alpha^{2 t-1} & \left(\alpha^{2 t-1}\right)^{2}\left(\alpha^{2 t-1}\right)^{3} \cdots & \left(\alpha^{2 t-1}\right)^{n-1}
\end{array}\right]
$$

- Example: Consider double- error correcting BCH Code of length 15.

$$
\begin{gathered}
15=2^{4}-1 \rightarrow m=4 \text { and from table 2.9: } \\
\varphi_{1}(x)=1+x+x^{4}, \varphi_{3}(x)=1+x+x^{2}+x^{3}+x^{4}
\end{gathered}
$$

So, $\mathrm{g}(\mathrm{x})=\varphi_{1}(\mathrm{x}) \varphi_{3}(\mathrm{x})=1+\mathrm{x}^{4}+\mathrm{x}^{6}+\mathrm{x}^{7}+\mathrm{x}^{8}$ and we have $n-k=8 \rightarrow k=15-8=7$

- So, this is the BCH Code $(15,7)$ with $d_{\text {min }}=5$, i.e., $\mathrm{t}=2$.


## Non-primitive BCH Codes

- Substituting $\alpha^{i}$,s, so we get:
$\mathrm{H}=\left[\begin{array}{l}100010011010111 \\ 010011010111100 \\ 001001101011110 \\ 000100110101111 \\ 100011000110001 \\ 000110001100011 \\ 001010010100101 \\ 011110111101111\end{array}\right]$


## - Example of a non-primitive BCH Code:

Consider $G F\left(2^{6}\right)$ and take $\beta=\alpha^{3}$. $\beta$ has order $n=21: \beta^{21}=\left(\alpha^{3}\right)^{21}=\alpha^{63}=1$

- Let $\mathrm{g}(\mathrm{x})$ be the minimal degree polynomial with roots: $\beta, \beta^{2}, \beta^{3}, \beta^{4}$
- $\beta, \beta^{2}$ and $\beta^{4}$ have the same minimal polynomial:

$$
\varphi_{1}(x)=1+x+x^{2}+x^{4}+x^{6}
$$

## Decoding of BCH Codes

and $B^{3}$ has: $\varphi_{3}(x)=1+x^{2}+x^{3}$. So $g(x)=\varphi_{1}(x) \varphi_{3}(x)=1+x+x^{4}+x^{5}+x^{7}+x^{8}+x^{9}$
It can be easily verified that $g(x)$ divides $x^{21}+1$. The code generated by $g(x)$ is a $(21,12)$ non-primitive BCH Code that corrects two errors.

- Decoding of BCH Codes:
- Let codeword $\underline{v}$ represented by code polynomial

$$
v(x)=v_{0}+v_{1} x+v_{2} x^{2}+\cdots+v_{n_{-} 1} x_{n_{-} 1}
$$

be the transmitted codeword.

- The received polynomial is:

$$
r(x)=r_{0}+r_{1} x+r_{2} x^{2}+\cdots+r_{n_{-} 1} x_{n_{-} 1}
$$

- Denoting the error polynomial by e(x), we have:

$$
r(x)=v(x)+e(x)
$$

- The syndrome is calculated multiplying $\underline{r}$ by $\mathrm{H}^{\mathrm{T}}$ :

$$
\underline{s}=\left(s_{1}, s_{2}, \ldots, s_{2 t}\right)=\underline{r} \cdot H^{T}
$$

## Decoding of BCH Codes

- This means that the $i-t h$ component of $\underline{\mathrm{s}}$ is:

$$
s_{i}=r\left(\alpha^{i}\right)=r_{0}+r_{1} \alpha^{i}+r_{2} \alpha^{2 i}+\cdots+r_{n-1} \alpha^{(n-1) i}
$$

for $i=1,2, \ldots, 2 t$.

- Let's divide $\mathrm{r}(\mathrm{x})$ by $\varphi_{i}(x)$, i.e., the minimal polynomial of $\alpha^{i}$ :

$$
r(x)=\alpha_{i}(x) \varphi_{i}(x)+b_{i}(x)
$$

- $\varphi_{i}\left(\alpha^{i}\right)=0$, therefore,

$$
S_{i}=r\left(\alpha^{i}\right)=b_{i}\left(\alpha^{i}\right)
$$

- Example: Consider ( 15,7 ) BCH Code. Let the received vector be (100000001000000). So, $r(x)=1+\mathrm{x}^{8}$. Let's find, $\underline{\mathrm{S}}=\left(s_{1}, s_{2}, s_{3}, s_{4}\right)$.The minimal polynomial for $\alpha, \alpha^{2}, \alpha^{4}$ is the same,

$$
\varphi_{1}(x)=\varphi_{2}(x)=\varphi_{4}(x)=1+x+x^{4}
$$

and for $\alpha^{3}$ we have,

$$
\varphi_{3}(x)=1+x+x^{2}+x^{3}+x^{4}
$$

## Decoding of BCH Codes

- Dividing $\mathrm{r}(\mathrm{x})=1+\mathrm{x}^{8}$ by $\varphi_{1}(x)$ we get,

$$
b_{1}(x)=x^{2}
$$

- Dividing $\mathrm{r}(\mathrm{x})$ by $\varphi_{3}(x)$, we get

$$
b_{3}(x)=1+x^{3}
$$

So,

$$
s_{1}=b_{1}(\alpha)=\alpha^{2}, \quad s_{2}=\alpha^{4}, \quad s_{4}=\alpha^{8}
$$

and

$$
s_{3}=b_{3}\left(\alpha^{3}\right)=1+\alpha^{9}=1+\alpha+\alpha^{3}=\alpha^{7}
$$

Therefore,

$$
\underline{S}=\left(\alpha^{2}, \alpha^{4}, \alpha^{7}, \alpha^{8}\right)
$$

## Decoding of BCH Codes

- Since

$$
V\left(\alpha^{i}\right)=0, \text { for } i=1,2, \ldots, 2 t
$$

we have

$$
S_{i}=r\left(\alpha^{i}\right)=v\left(\alpha^{i}\right)+e\left(\alpha^{i}\right)=e\left(\alpha^{i}\right)
$$

- Now, assume that we have $v$ errors at locations $j_{1}, j_{2}, \ldots, j_{\gamma}$. That is,

$$
e(x)=x^{j_{1}}+x^{j_{2}}+\cdots+x^{v}
$$

- Then we have,

$$
\begin{gathered}
S_{1}=\alpha^{j_{1}}+\alpha^{j_{2}}+\ldots+\alpha^{j_{v}} \\
S_{2}=\left(\alpha^{j_{1}}\right)^{2}+\left(\alpha^{j_{2}}\right)^{2}+\cdots+\left(\alpha^{j_{v}}\right)^{2} \\
\vdots \\
S_{2 t}=\left(\alpha^{j_{1}}\right)^{2 t}+\left(\alpha^{j_{2}}\right)^{2 t}+\cdots+\left(\alpha^{j_{v}}\right)^{2 t}
\end{gathered}
$$

## Decoding of BCH Codes

Let $\beta_{1}=e^{j_{1}} \beta_{2}=e^{j_{2}}, \ldots, \beta_{\gamma}=e^{j_{\gamma}}, \beta_{1,}, \beta_{2}, \ldots, \beta_{\gamma}$ are called error location numbers. Then we have:

$$
\begin{gathered}
\mathrm{S}_{1}=\beta_{1}+\beta_{2}+\ldots+\beta_{v} \\
\mathrm{~S}_{2}=\beta_{1}{ }^{2}+\beta_{2}{ }^{2}+\ldots+\beta_{v}{ }^{2} \\
\vdots \\
\mathrm{~S}_{2 \mathrm{t}}=\beta_{1}{ }^{2 \mathrm{t}}+\beta_{2}{ }^{2 \mathrm{t}}+\ldots+\beta_{v}{ }^{2 \mathrm{t}}
\end{gathered}
$$

These 2 t equations are symmetric function of $\beta_{1}, \beta_{2}, \ldots, \beta_{v}$

- Define the following polynomial

$$
\sigma(x)=\left(1+\beta_{1} x\right)\left(1+\beta_{2} x\right)\left(1+\beta_{3} x\right) \ldots\left(1+\beta_{v} x\right)
$$

This is called the error locator polynomial and has $\beta_{1}^{-1} \beta_{2}^{-1} \ldots \beta_{v}^{-1}$ as its roots. $\sigma(\mathrm{X})$ can also be represented as:

$$
\sigma(x)=\sigma_{0}+\sigma_{1} x+\sigma_{2} x^{2}+\cdots+\sigma_{v} x^{v}
$$

## Decoding of BCH Codes

It is clear that:

$$
\begin{gathered}
\sigma_{0}=1 \\
\sigma_{1=}=\mathrm{B}_{1}+\mathrm{B}_{2}+\ldots+\beta_{v} \\
\sigma_{2=} \mathrm{B}_{1} \mathrm{~B}_{2}+\mathrm{B}_{2} \mathrm{~B}_{3}+\ldots+\beta_{v-1} \beta_{v} \\
\vdots \\
\sigma_{\gamma=} B_{1} B_{2} \ldots \beta_{v}
\end{gathered}
$$

- $\sigma_{i}, \mathrm{~s}$ can be shown to be related to syndromes as follows:

$$
\begin{gathered}
s_{1}+\sigma_{1}=0 \\
s_{2}+\sigma_{1} s_{1}+2 \sigma_{2}=0 \\
s_{3}+\sigma_{1} s_{2} \sigma_{2} s_{1}+3 s_{3}=0 \\
\vdots \\
s_{v}+\sigma_{1} s_{v-1+\cdots+} \sigma_{v-1} s_{1}+v \sigma_{v}=0 \\
s_{v+1}+\sigma_{1} s_{v+\cdots+} \sigma_{v-1} s_{2}+v s_{1}=0
\end{gathered}
$$

- These are called Newton identities.
- For the binary case

$$
i \sigma_{i}= \begin{cases}\sigma_{i} & \text { for odd } i \\ 0 & \text { for even } i\end{cases}
$$

## Berlekamp Algorithm

- Berlekamp Algorithm is an Iterative Algorithm for finding Error-Location Polynomial:
This algorithm tries to generate polynomials of degree $1,2, .$. that has $\beta_{1}, \beta_{2} \ldots$ as it roots.
- First we define $\sigma^{(1)}(x)$ that satisfies the first Newton equality: $\sigma^{(1)}(\mathrm{x})=1+\mathrm{S}_{1} \mathrm{x}$ Since $\mathrm{S}_{1}+\sigma_{1}=0 \rightarrow \sigma_{1}=\mathrm{S}_{1}$
- Then we check whether $\sigma^{(1)}(x)$ satisfies the second Newton equality or not. If it satisfies we let $\sigma^{(2)}(x)=\sigma^{(1)}(x)$ otherwise we add another term to $\sigma^{(1)}(x)$ to form $\sigma^{(2)}(x)$ that satisfies the first and second equalities.
- Then for $\sigma^{(3)}(x)$ : if $\sigma^{(2)}(x)$ satisfies the third equality we let $\sigma^{(3)}(x)=$ $\sigma^{(2)}(x)$ otherwise add a correction term that makes $\sigma^{(3)}(x)$ satisfy the first three equalities.
- We continue this iterative approach until we get $\sigma^{(2 t)}(x)$ and set $\sigma(x)=\sigma^{(2 t)}(x)$.
- Now let's see how we can go from one stage say $\mu$ to $\mu+1$.


## Berlekamp Algorithm

- Assume that at stage $\mu$, the polynomial is

$$
\sigma^{(\mu)}(x)=1+\sigma_{1}^{(\mu)} x+\sigma_{2}^{(\mu)} x^{2}+\ldots+\sigma_{L_{\mu}}^{(\mu)} x^{L_{\mu}}
$$

- If $\sigma^{(\mu)}(x)$ satisfies also $(\mu+1)$ st equality then, $\mathrm{S}_{\mu+1}$ should be

$$
\sigma_{1}^{(\mu)} s_{\mu}+\sigma_{2}^{(\mu)} s_{\mu-1}+\ldots+\sigma_{L_{\mu}}^{(\mu)} s_{\mu+1-L_{\mu}}
$$

- We compare this with actual $s_{\mu+1}$. That is why we add this to $S_{\mu+1}$ and check whether we get zero or not Let the sum be denoted by $d_{\mu}$ and call it discrepancy.

$$
d_{\mu}=s_{\mu+1}+\sigma_{1}^{(\mu)} s_{\mu}+\sigma_{2}^{(\mu)} s_{\mu-1}+\ldots+\sigma_{L_{\mu}}^{(\mu)} s_{\mu+1-L_{\mu}}
$$

- If this is zero, then $\sigma^{(\mu)}(x)$ also satisfies the $\mu+1$-st equality and therefore,

$$
\sigma^{(\mu+1)}(x)=\sigma^{(\mu)}(x)
$$

- But if $d_{\mu} \neq 0$, then $\sigma^{(\mu)}(x)$ does not satisfy the $\mu+1$-st equality.


## Berlekamp Algorithm

- Note that

and

$$
d_{\rho}=\sum_{i=0}^{L \rho} \sigma_{i}^{(\rho)} s_{\rho_{+} 1_{-} i}
$$

$$
\sigma^{(\rho)}(\mathrm{x})=1+\sigma_{1}^{(\rho)}{ }_{\mathrm{x}}+\sigma_{2}^{(\rho)} \mathrm{x}^{2}+\ldots+\sigma_{L \rho}^{(\rho)} x^{L \rho}
$$

- Let's form $\sigma^{(\mu+1)}(x)$ as:

$$
\sigma^{(\mu+1)}(x)=\sigma^{(\mu)}(x)+A X^{\mu-\rho} \sigma^{(\rho)}(x)
$$

- Then

$$
d_{\mu}^{\prime}=\sum_{i=0}^{L \mu} \sigma_{i}^{(\mu)} \mathrm{S}_{\mu+1-\mathrm{i}}+\sum_{i=0}^{L \rho} \sigma_{i}^{(\rho)} \mathrm{S}_{\mu-\rho+1-\mathrm{i}}
$$

or

$$
\begin{gathered}
d_{\mu}^{\prime}=d_{\mu}+A d_{\rho} \\
A=d_{\mu} / d_{\rho}
\end{gathered}
$$

## Summary of Berlekamp Algorithm

- In summary, Berlekamp algorithm is as follows:
- Initialization: start with first two rows according to the following table:

| $\boldsymbol{\mu}$ | $\sigma^{(\mu)}(\boldsymbol{X})$ | $\boldsymbol{d}_{\mu}$ | $\boldsymbol{I}_{\boldsymbol{\mu}}$ | $\mu-l_{\mu}$ |
| :---: | :---: | :---: | :---: | :---: |
| -1 | 1 | 1 | 0 | -1 |
| 0 1 | 1 | $S_{1}$ | 0 | 0 |
| 1 |  |  |  |  |
| ! |  |  |  |  |
| $2 t$ |  |  |  |  |

- Iteration: For each $\mu$ form $d_{\mu}=s_{\mu+1}+\sigma_{1}^{(\mu)} s_{\mu}+\cdots+\sigma_{L \mu}^{(\mu)} x$

Where $L_{\mu}$ is the degree of $\sigma^{(\mu)}(x)$

## Summary of Berlekamp Algorithm

1) If $d_{\mu}=0$ then $\sigma^{(\mu+1)}(x)=\sigma^{(\mu)}(x)$
2) If $d_{\mu} \neq 0$ then:

$$
\sigma^{(\mu+1)}(x)=\sigma^{(\mu)}(x)+d_{\mu} d_{\rho}^{-1} x^{\mu-\rho} \sigma^{(\rho)}(x)
$$

Where $\rho$ is the row (the stage) where $d_{\rho} \neq 0$ and is closest to $\mu$, i.e., $\mu-\rho$ is the smallest

- Termination:
- Continue until you find $\sigma^{(2 t)}(x)$ and let:

$$
\sigma(x)=\sigma^{(2 t)}(x)
$$

## Example

- Consider the $(15,5)$ code we saw previously assume that,

$$
\begin{aligned}
\mathrm{v}= & (0,0,0,0,0,0,0,0,0,0,0,0,0,0,0) \text { is transmitted } \\
& \text { and } \mathrm{r}=(000101000000100) \text { is received. }
\end{aligned}
$$

Then $r(x)=x^{3}+x^{5}+x^{12}$.

- The minimal polynomial for $\alpha, \alpha^{2}$ and $\alpha^{4}$ is

$$
\varphi_{1}(x)=\varphi_{2}(x)=\varphi_{4}(x)=1+x+x^{4}
$$

(For $\alpha^{3}$ and $\alpha^{6}$

$$
\varphi_{3}(x)=\varphi_{6}(x)=1+x+x^{2}+x^{3}+x^{4}
$$

- For $\alpha^{5}$,

$$
\varphi_{5}(x)=1+x+x^{2}
$$

- Dividing $\mathrm{r}(\mathrm{x})$ by $\varphi_{1}(x)$, we get

$$
b_{1}(x)=1
$$

- Dividing $\mathrm{r}(\mathrm{x})$ by $\varphi_{3}(x)$, we get

$$
b_{3}(x)=1+x^{2}+x^{3}
$$

- And dividing by $\varphi_{5}(x)$,

$$
b_{5}(x)=x^{2}
$$

## Example

So:

$$
\begin{gathered}
s_{1}=s_{2}=s_{4}=1 \\
s_{3}=1+\alpha^{6}+\alpha^{9}=\alpha^{10} \\
s_{6}=1+\alpha^{12}+\alpha^{18}=\alpha^{5} \\
s_{5}=\alpha^{10}
\end{gathered}
$$

Using Berlekamp method, we get $\sigma(x)=\alpha^{(6)}(x)=1+x+\alpha^{5} x$.

| $\boldsymbol{\mu}$ | $\boldsymbol{\sigma}^{(\mu)}(\boldsymbol{X})$ | $\boldsymbol{d}_{\boldsymbol{\mu}}$ | $\boldsymbol{l}_{\boldsymbol{\mu}}$ | $\boldsymbol{\mu}-\boldsymbol{l}_{\boldsymbol{\mu}}$ |
| ---: | :--- | :--- | :--- | :--- |
| -1 | 1 | 1 | 0 | -1 |
| 0 | 1 | 1 | 0 | 0 |
| 1 | $1+X$ | 0 | 1 | 0 (take $\rho=-1$ ) |
| 2 | $1+X$ | $\alpha^{5}$ | 1 | 1 |
| 3 | $1+X+\alpha^{5} X^{2}$ | 0 | 2 | 1 (take $\rho=0$ ) |
| 4 | $1+X+\alpha^{5} X^{2}$ | $\alpha^{10}$ | 2 | 2 |
| 5 | $1+X+\alpha^{5} X^{3}$ | 0 | 3 | 2 (take $\rho=2$ ) |
| 6 | $1+X+\alpha^{5} X^{3}$ | - | - | - |

## Example

- We can verify that $\alpha^{3}, \alpha^{10}$ and $\alpha^{12}$ are the roots of $\sigma(\mathrm{x})$.

$$
\begin{aligned}
& \left(\alpha^{3}\right)^{-1}=\alpha^{12} \\
& \left(\alpha^{10}\right)^{-1}=\alpha^{5}
\end{aligned}
$$

and

$$
\left(\alpha^{12}\right)^{-1}=\alpha^{3}
$$

- So:

$$
e(x)=x^{3}+x^{5}+x^{12}
$$

## Error Correction Procedure

1) Calculate syndrome.
2) Form error- location polynomial $\sigma(x)$
3) Solve $\sigma(x)$ to get error locations (Chien Search)

- Chien Search:

1) Load $\sigma_{1,} \sigma_{2, \ldots, \ldots} \sigma_{2 t}$ in 2 t registers.
(If $\sigma(\mathrm{x})$ has degree less than 2 t , i.e., $\mu<2 t$ then $\sigma_{\mu+1}=\sigma_{\mu+2}=\cdots=\sigma_{2 t}=0$ )
2) The multipliers multiply $\sigma_{i}$ by $\alpha^{i}$ and the circuit generates

$$
\sigma_{1} \alpha+\sigma_{2} \alpha^{2}+\cdots+\sigma_{\mu} \alpha^{\mu}
$$

- If $\alpha$ is a root of $\sigma(x)$ then

$$
1+\sigma_{1} \alpha+\sigma_{2} \alpha^{2}+\cdots+\sigma_{\mu} \alpha^{\mu}=0
$$

## Chien Search



Load $\sigma_{1,} \sigma_{2, \ldots, \sigma_{2 t}}$ in 2 t registers.
(If $\sigma(\mathrm{x})$ has degree less than 2t, i.e., $\mu<2 t$ then $\sigma_{\mu+1}=\sigma_{\mu+2}=\cdots=\sigma_{2 t}=0$ )
The multipliers multiply $\sigma_{i}$ by $\alpha^{i}$ and the circuit generates

$$
\sigma_{1} \alpha+\sigma_{2} \alpha^{2}+\cdots+\sigma_{\mu} \alpha^{\mu}
$$

- If $\alpha$ is a root of $\sigma(x)$ then

$$
1+\sigma_{1} \alpha+\sigma_{2} \alpha^{2}+\cdots+\sigma_{\mu} \alpha^{\mu}=0
$$

## Error Correction Procedure

- Or the output of A is 1 .
- So if output of A is 1 then $\alpha$ is a root and $\alpha^{-1}=\alpha^{n-1}$ is error location and $r_{n-1}$ should be corrected.
- Multipliers are clocked so we get

$$
\alpha^{2},\left(\alpha^{2}\right)^{2}, \ldots,\left(\alpha^{2}\right)^{\mu}
$$

Or the output of A is

$$
\sigma_{1} \alpha^{2}+\sigma_{2}\left(\alpha^{2}\right)^{2}+\cdots \sigma_{\mu}\left(\alpha^{2}\right)^{\mu}
$$

If this is $1, r_{n-2}$ should be corrected and so on for $3, . ., \nu$.

