

Q.1: (Sol/hints)

Consider Early effect for both M1 and M2 (for best accuracy)

```
> VA:=25.;
VA := 25.

> Vo:=3.;
Vo := 3.

> Kn:=100.;
Kn := 100.

> L:=1.;
L := 1.

> W1:=2.;
W1 := 2.

> W2:=4.;
W2 := 4.

> Vth:=0.6;
Vth := .6

> Io:=50.;
Io := 50.

> y1:=Io-Kn*(W2/(2*L))*Vov^2*(1+Vo/VA);
y1 := 50. - 224.0000000 Vov^2

> solve(y1,Vov);
-.4724555913, .4724555913

> Vov1:=.4724555913;
Vov1 := .4724555913

> VG2:=Vth+Vov1;
VG2 := 1.072455591

> VGS1:=VG2;
VGS1 := 1.072455591

> VDS1:=VGS1;
VDS1 := 1.072455591

> x1:=1+VDS1/VA;
x1 := 1.042898224

> x2:=1+Vo/VA;
x2 := 1.120000000

> IREF:=(x1/x2)*(Io/2);
IREF := 23.27897822

> R:=1E6*(5-VG2)/IREF;
R := 168716.3574

R=168.72 k ohms
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Q.2: (Sol/hint)

We need to add a series resistance with the internal resistance of the transistor to divide the 50mV into 3mV +47mV.

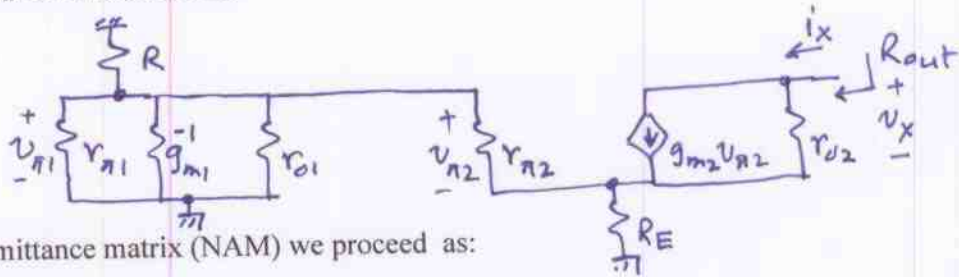
For each BJT, 're' =25/5 =50 ohms. If RE is added to each emitter leg, (RE+re)/re =25/3. This gives RE

Show the schematic with RE included in the emitter leg. An alternative could be to add (beta+1)\*RE in series with the base of each BJT.

The voltage gain will be the gain of a single CE amplifier with emitter resistance RE. It is approximately =-alpha\*RC/(re+RE), where alpha =beta/(beta+1). Note the minus sign!

Q.3: (Sol/hint)

The Rout to be investigated is a 'small signal' (ac) quantity. So the ac equivalent circuit for the BJTs will have to be drawn. The circuit is:



For the nodal admittance matrix (NAM) we proceed as:

$$\begin{bmatrix} g_{o2} & -g_{o2} & 0 \\ -g_{o2} & g_E + g_{o2} + g_{\pi 2} & -g_{\pi 2} \\ 0 & -g_{\pi 2} & g_{\pi 2} + g_{o1} + g_{\pi 1} + g_R \end{bmatrix} \begin{bmatrix} V_x \\ V_2 \\ V_3 \end{bmatrix} = \begin{bmatrix} i_x - g_{m2} v_{\pi 2} \\ g_{m2} v_{\pi 2} \\ -g_{m1} v_{\pi 1} \end{bmatrix}$$

Noting  $v_{\pi 2} = V_3 - V_2$ ,  $v_{\pi 1} = V_3$ , we get the NAM

$$\begin{bmatrix} g_{o2} & -g_{m2} - g_{o2} & g_{m2} \\ -g_{o2} & g_E + g_{o2} + g_{\pi 2} + g_{m2} & -g_{\pi 2} - g_{m2} \\ 0 & -g_{\pi 2} & g_{\pi 2} + g_{o1} + g_{\pi 1} + g_R + g_{m1} \end{bmatrix} \begin{bmatrix} V_x \\ V_2 \\ V_3 \end{bmatrix} = \begin{bmatrix} i_x \\ 0 \\ 0 \end{bmatrix}$$

The Rout is =Vx/ix. Solve by Krammer's rule:

$$V_x = \frac{\begin{vmatrix} i_x & -g_{m2} - g_{o2} & g_{m2} \\ 0 & g_E + g_{o2} + g_{\pi 2} + g_{m2} & -g_{\pi 2} - g_{m2} \\ 0 & -g_{\pi 2} & g_{\pi 2} + g_{o1} + g_{\pi 1} + g_R + g_{m1} \end{vmatrix}}{\begin{vmatrix} g_{o2} & -g_{m2} - g_{o2} & g_{m2} \\ -g_{o2} & g_E + g_{o2} + g_{\pi 2} + g_{m2} & -g_{\pi 2} - g_{m2} \\ 0 & -g_{\pi 2} & g_{\pi 2} + g_{o1} + g_{\pi 1} + g_R + g_{m1} \end{vmatrix}}$$

> x1 := gE+go2+gp2+gm2 ;

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x1 := gE + go2 + gp2 + gm2
> x2 := gp2 + go1 + gp1 + gR + gm1;
x2 := gp2 + go1 + gp1 + gR + gm1
> x3 := -gp2 - gm2;
x3 := -gp2 - gm2
> x4 := -gp2;
x4 := -gp2
> z1 := x1 * x2 - x3 * x4;
z1 := (gE + go2 + gp2 + gm2) (gp2 + go1 + gp1 + gR + gm1) + (-gp2 - gm2) gp2
> simplify(z1);
gE gp2 + gE go1 + gE gp1 + gE gR + gE gm1 + go2 gp2 + go2 go1 + go2 gp1 + go2 gR
+ go2 gm1 + gp2 go1 + gp2 gp1 + gp2 gR + gp2 gm1 + gm2 go1 + gm2 gp1 + gm2 gR + gm2 gm1
> x5 := -(gm2 + go2) * x2 + gp2 * gm2;
x5 := -(gm2 + go2) (gp2 + go1 + gp1 + gR + gm1) + gm2 gp2
> simplify(x5);
-gm2 go1 - gm2 gp1 - gm2 gR - gm2 gm1 - go2 gp2 - go2 go1 - go2 gp1 - go2 gR - go2 gm1
> x6 := -gm2 * go1 - gm2 * gp1 - gm2 * gR - gm2 * gm1 - go2 * gp2 - go2 * go1 - go2 * gp1 - go2 * gR
-gm2 * gm1;
x6 :=
-gm2 go1 - gm2 gp1 - gm2 gR - gm2 gm1 - go2 gp2 - go2 go1 - go2 gp1 - go2 gR - go2 gm1
> z2 := go2 * (z1 + x6);
z2 := go2 ((gE + go2 + gp2 + gm2) (gp2 + go1 + gp1 + gR + gm1) + (-gp2 - gm2) gp2
-gm2 go1 - gm2 gp1 - gm2 gR - gm2 gm1 - go2 gp2 - go2 go1 - go2 gp1 - go2 gR - go2 gm1
)
> simplify(z2);
go2 (gE gp2 + gE go1 + gE gp1 + gE gR + gE gm1 + gp2 go1 + gp2 gp1 + gp2 gR + gp2 gm1)
> z3 := go2 * (gE * gp2 + gE * go1 + gE * gp1 + gE * gR + gE * gm1 + gp2 * go1 + gp2 * gp1 + gp2 * gR +
gp2 * gm1);
z3 :=
go2 (gE gp2 + gE go1 + gE gp1 + gE gR + gE gm1 + gp2 go1 + gp2 gp1 + gp2 gR + gp2 gm1)
Rout = Vx / ix. Thus:
> Rout := z1 / z3;
Rout :=
(gE + go2 + gp2 + gm2) (gp2 + go1 + gp1 + gR + gm1) + (-gp2 - gm2) gp2
go2 (gE gp2 + gE go1 + gE gp1 + gE gR + gE gm1 + gp2 go1 + gp2 gp1 + gp2 gR + gp2 gm1)
>

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Q.4 (Sol/hint)

We can use the half-circuit concept by dividing  $R_s$  into two halves with an 'ac' ground at the center point. Thus each half will be like a CS MOS amplifier with a resistance  $R_s/2$  at the source terminal.

From the result of a CS amplifier with  $R_x (=R_s/2)$  at the source end, the voltage gain is:

$$G = \frac{g_m R_D}{1 + g_m (R_s / 2)}$$

Thus  $A = g_m R_D$ . This is the answer to part (a) of the question.

For part(b) of the question, we solve:  $g_m R_D / 2 = \frac{g_m R_D}{1 + g_m R_s / 2}$

>  $yx := gm * RD / 2 - (gm * RD) / (1 + gm * Rs / 2) ;$

$$yx := \frac{1}{2} gm RD - \frac{gm RD}{1 + \frac{1}{2} gm Rs}$$

>  $solve(yx, Rs) ;$

$$2 \frac{1}{gm}$$

[  $Rs = 2/gm$  is the solution.