

Sol/hint for Q.1 (ELEC 311-Summer 2013)

```
> id1:=1E-3*4*(2.5-V1-1)^2;
id1 := .004 (1.5 - V1)2
> id2:=1E-3*4*(-V2-1)^2;
id2 := .004 (-V2 - 1)2
> y1:=v2-id2*1200+2.5;
y1 := V2 - 4.800 (-V2 - 1)2 + 2.5
> solve(y1,v2);
-1.464472669, -3271939973
```

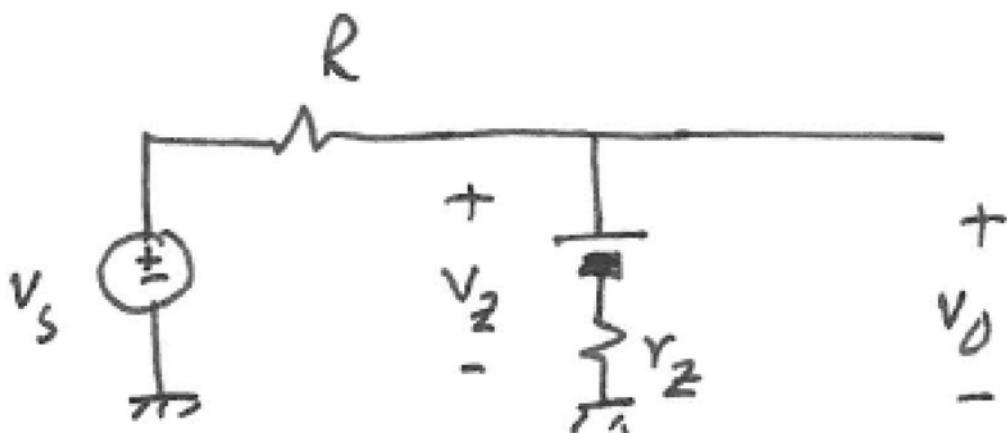
For MN2, VGS must be > VTH. But VGS|MN2=-V2, and VTH=1V.
So -V2>1V, V2<-1, accepted V2=-1.46 V (approx).

```
> V2:=-1.46;
V2 := -1.46
> id2:=4E-3*(-V2-1)^2;
id2 := .0008464
> id1:=id2;
id1 := .0008464
> y2:=id1-4E-3*(1.5-V1)^2;
y2 := .0008464 - .004 (1.5 - V1)2
> solve(y2,v1);
1.040000000, 1.960000000
```

Since VGS of MN1 has to be >VTH, our choice for V1 will be V1=1.04V
So solutions are V1=1.04V, V2=-1.46

Sol/hint for Q.2

(a) The equivalent circuit is (hand-drawing)



(b) **Vz:=Vzo+Iz*rz;**

```

Vz := Vzo + Iz * rz
> Vz := 6.8;
Vz := 6.8

> Iz := 5E-3;
Iz := .005

> rz := 20;
rz := 20

> y := Vz - Vzo - Iz * rz;
y := 6.700 - Vzo

> solve(y, Vzo);
6.700000000

> Vzo := 6.7;
Vzo := 6.7

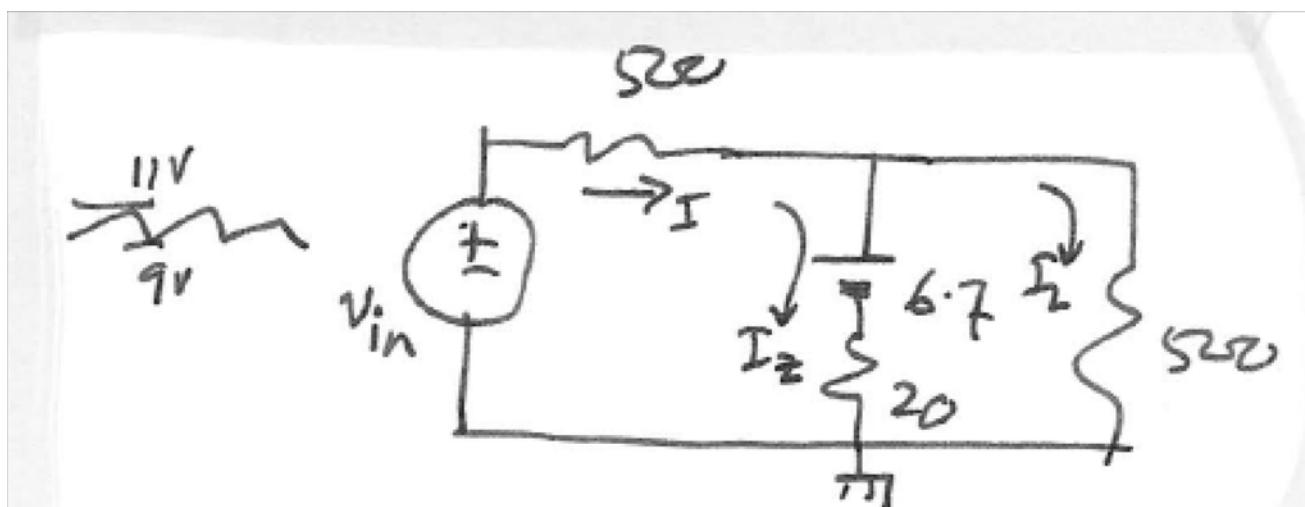
Line regulation is:  $\frac{\Delta V_o}{\Delta V_i} = \frac{\Delta V_z}{\Delta V_i} = \frac{r_z}{r_z + R_{line}}$ 
> Rline := 500;
Rline := 500

> LR := rz / (rz + Rline);
LR :=  $\frac{1}{26}$ 

```

Line regulation is 1/26, i.e., 38.46 mV/V

(c) With 500 ohms as load, the equivalent circuit is (hand drawing):



If we solve this circuit for I_z , it will turn out that I_z is negative, i.e., the Zener is OFF. Thus

```
> Iline := (Vin - Vload) / Rline;
```

$$I_{line} := \frac{1}{500} V_{in} - \frac{1}{500} V_{load}$$

> **Vload:=Vzener;**

$$V_{load} := V_{zener}$$

$$V_{zener} = I_z * r_z + V_{zo}$$

$$I_z + I_{load} = V_{in}/500 - V_{zener}/500, \text{ while } I_{load} = V_{zener}/500;$$

Substituting the values $V_{in}=10V$, $V_{zo}=6.7V$, $r_z=20$, solving for I_z will reveal $I_z < 0$.

With the Zener OFF, the output voltage is $V_{in}*(500)/(500+500) = 5V$ for nominal $V_{in} = 10V$

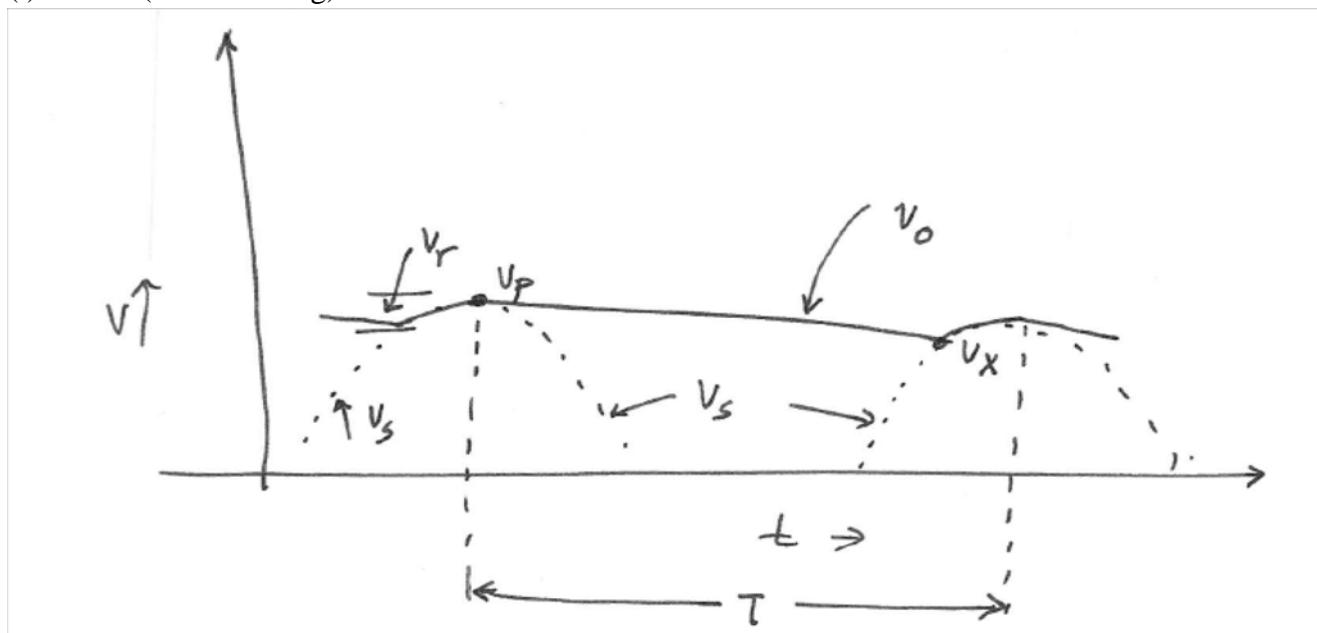
Q.3(a):

(i) DC component = $V_m/\pi = 60/3.142 = 19.1$ volts

(ii) PIV = $0 - (-60) = 60$ volts

Q.3(b):

(i) Sketch (hand drawing)



(ii) The time constant $CR = 100,000 * 1E-5 = 1$ sec. The supply period is : $2\pi/(120\pi) = 1/60 = 16.7$ m sec. So $CR \gg$ supply period.

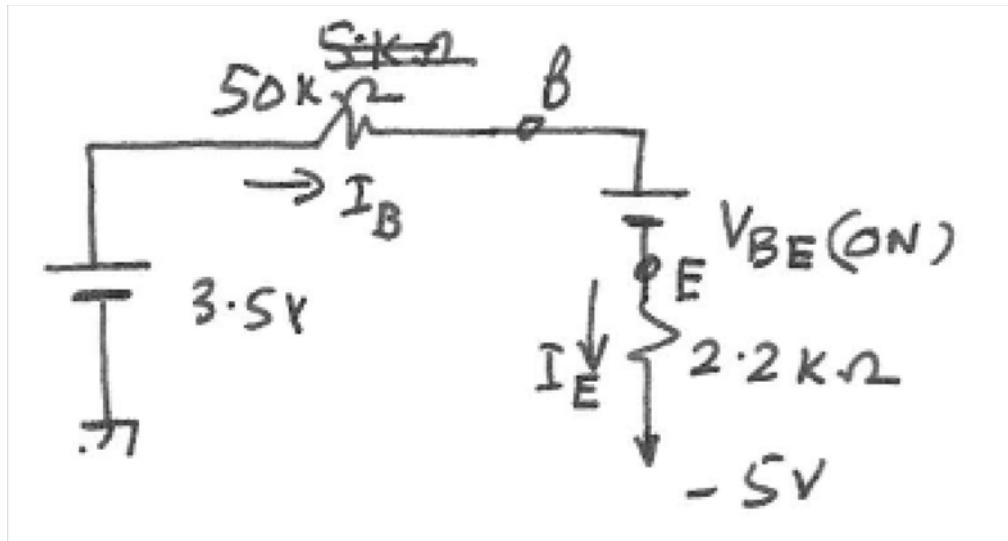
In each cycle, V_o starts from $V_p = 60V$, and decays at the time-constant rate of 1 sec for (approx.) a time of one whole period, i.e., 16.7 m sec. The decayed value is (discharge equation for a simple RC circuit): $60 \exp(-16.7E-3/1) = 60 * (1 - 0.0167) = 59V$ (approx.)

The DC value at output is : $(1/2)[60+59] = 59.5V$

The ripple is $60-59=1V$ pk-pk

(iii) To halve the ripple, we increase the capacitance to double the value, i.e., 20 micro farads.

Q.4: First we need to find I_{dc} , i.e., DC analysis. Consider the base-emitter loop for the DC voltages (hand drawing)

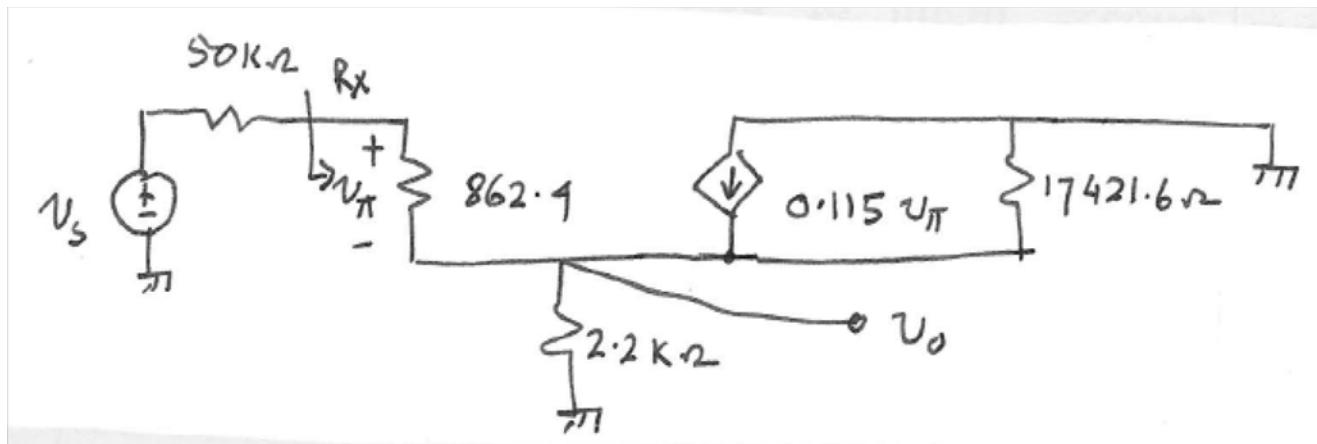


```

y:=3.5-(IE/100)*50E3-0.7-IE*2200+5;
y := 7.8 - 2700.000000 IE
> solve (y,IE);
.002888888889
> IE:=2.9E-3;
IE := .0029
> IC:=IE*beta/(beta+1);
IC := .0029  $\frac{\beta}{\beta + 1}$ 
> beta:=99;
beta := 99
> ICnow:=IC;
ICnow := .002871000000
> Ic=2.87E-3;
Ic = .00287
> gm:=0.00287/.025;
gm := .1148000000
> rpi:=beta/gm;
rpi := 862.3693380
> rpi := 862.3693380;
rpi := 862.3693380
> ro:=50/.00287;
ro := 17421.60279
>

```

The ac equivalent circuit is (hand drawing)



> RE:=2200;

$$RE := 2200$$

> Rx:=862.4+100*(RE*ro)/(RE+ro);

$$Rx := 196195.7097$$

> vb:=Rx*vs/(Rx+50000);

$$vb := .7969095397 \text{ vs}$$

> RL:=100*(RE*ro)/(RE+ro);

$$RL := 195333.3097$$

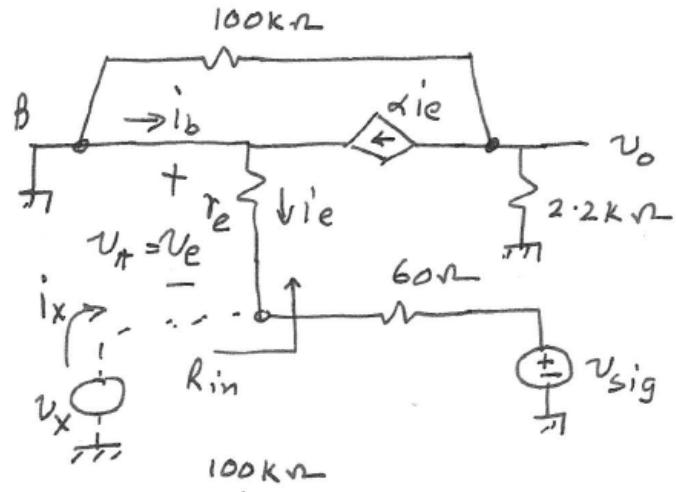
> vo:=vb*RL/Rx;

$$vo := .7934066353 \text{ vs}$$

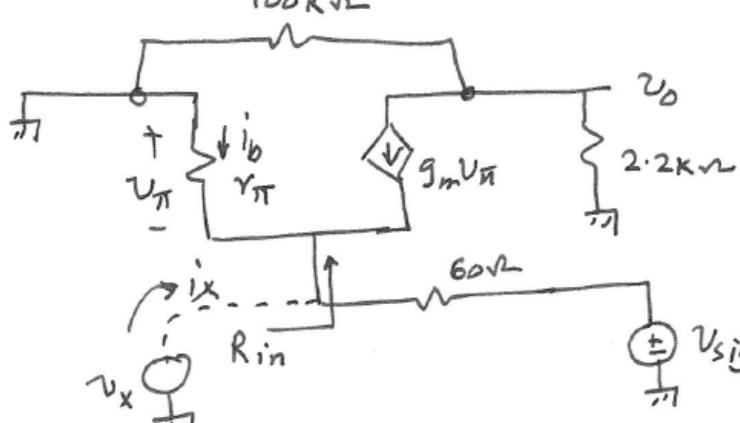
Voltage gain is $vo/vs = 0.79$

Q.5:

- (a) Consider the ac equivalent circuit (T-model and PI-model). Using the dummy source v_x and the dummy current i_x , we get $R_{in} = \frac{v_x}{i_x} = r_e = \frac{r_\pi}{1 + \beta}$



$$\begin{aligned}
 -i_x - i_e &= 0 \\
 i_x &= -i_e \\
 v_x &= -v_e = -i_e r_e \\
 R_{in} &= \frac{v_x}{i_x} = r_e
 \end{aligned}$$



$$\begin{aligned}
 -i_x - i_b - g_m v_\pi &= 0 \\
 i_x &= -i_b - g_m i_b r_\pi \\
 &= -(1+\beta) i_b \\
 i_b &= v_\pi / r_\pi \\
 v_x &= -v_\pi \\
 \frac{v_x}{i_x} &= R_{in} = \frac{r_\pi}{1+\beta}
 \end{aligned}$$

From the given schematic $0.5\text{mA} = \text{IE} = \text{IC} + \text{IB}$. So $\text{IC} = (49/50) * 0.5\text{mA} = 0.49\text{ mA}$

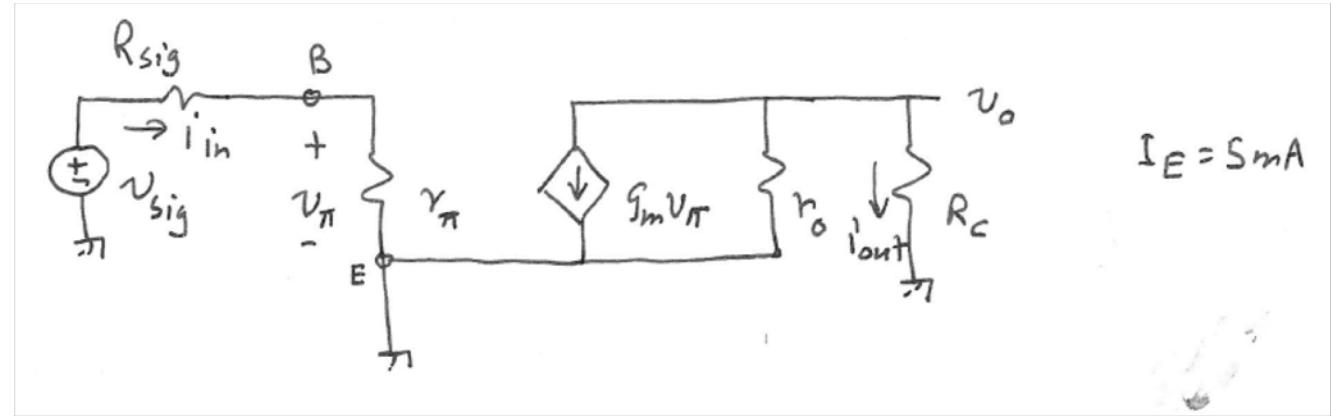
$$gm = 0.0196 \text{ mho}; r_\pi = \frac{\beta}{g_m} = 2500$$

$R_{in} = 50 \text{ ohms}$ (i.e., Thermal voltage divided by IE).

(b) From the equivalent circuit $v_e = v_\pi = \frac{50}{50+60} v_{sig}$ has to be $< V_T$ (i.e., 25 mV)

$$\text{Then } v_s = v_{sig} \text{ has to be } < \frac{25 \times 110}{50} \text{ mV} = 55 \text{ mV}$$

Q.6: The equivalent circuit is (hand drawing):



IE:=.005;

IE := .005

> **beta:=100;**

β := 100

> **IC:=IE*beta/(1+beta);**

IC := .004950495050

> **VA:=50;**

VA := 50

> **ro:=VA/IC;**

ro := 10100.00000

> **gm:=IC/.025;**

gm := .1980198020

> **rpi:=beta/gm;**

rpi := 504.9999999

Put the above values in the equivalent circuit.

(a) The voltage gain is: $\frac{v_o}{v_{sig}} = -g_m \frac{r_o R_C}{r_o + R_C} \times \frac{r_\pi}{r_\pi + R_{sig}}$

Substituting the values

> **x1:=(ro*1000)/(ro+1000);**

x1 := 909.9099099

> **x2:=rpi/(rpi+10000);**

x2 := .04807234649

> **gain:=-gm*x1*x2;**

gain := -8.661684053

Voltage gain is -8.66 V/V

(b)

> **iout:=vo/1000;**

iout := \frac{1}{1000} vo

```

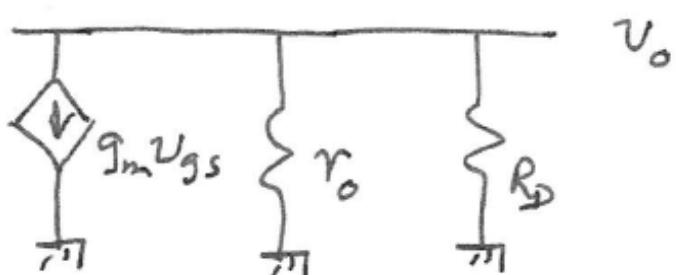
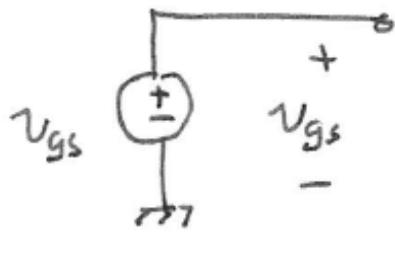
> iin:=vsig/(rpi+10000);
          iin := .00009519276535 vsig
> igain:=iout/iin;
          igain := 10.50500000  $\frac{vo}{vsig}$ 
> i_gain:=10.505*gain;
          i_gain := -90.99099098
Small signal current gain is -90.99 A/A

Q.7: Using several of the the given data
> K:=.005;
          K := .005
> WLratio:=2;
          WLratio := 2
> Vov:=0.7-0.5;
          Vov := .2
> gm:=K*WLratio*Vov;
          gm := .0020

```

(a) The ac transconductance is 0.002 mho

(b) The ac equivalent circuit is (hand drawing):



The DC current is $(ID) = K \frac{W}{2L} V_{ov}^2 = .005 \times \frac{2}{2} \times 0.2^2$ (assuming saturation region of operation)

```

> ID:=K*(WLratio/2)*Vov^2;
          ID := .00020000000000

```

```

> VD:=2-ID*5000;
          VD := 1.0000000000

```

```

> VDS:=VD;
          VDS := 1.0000000000

```

VDS=1V is > Vov (=0.2V), so the operation in saturation region is confirmed.

```

> VAX:=50;

```

```

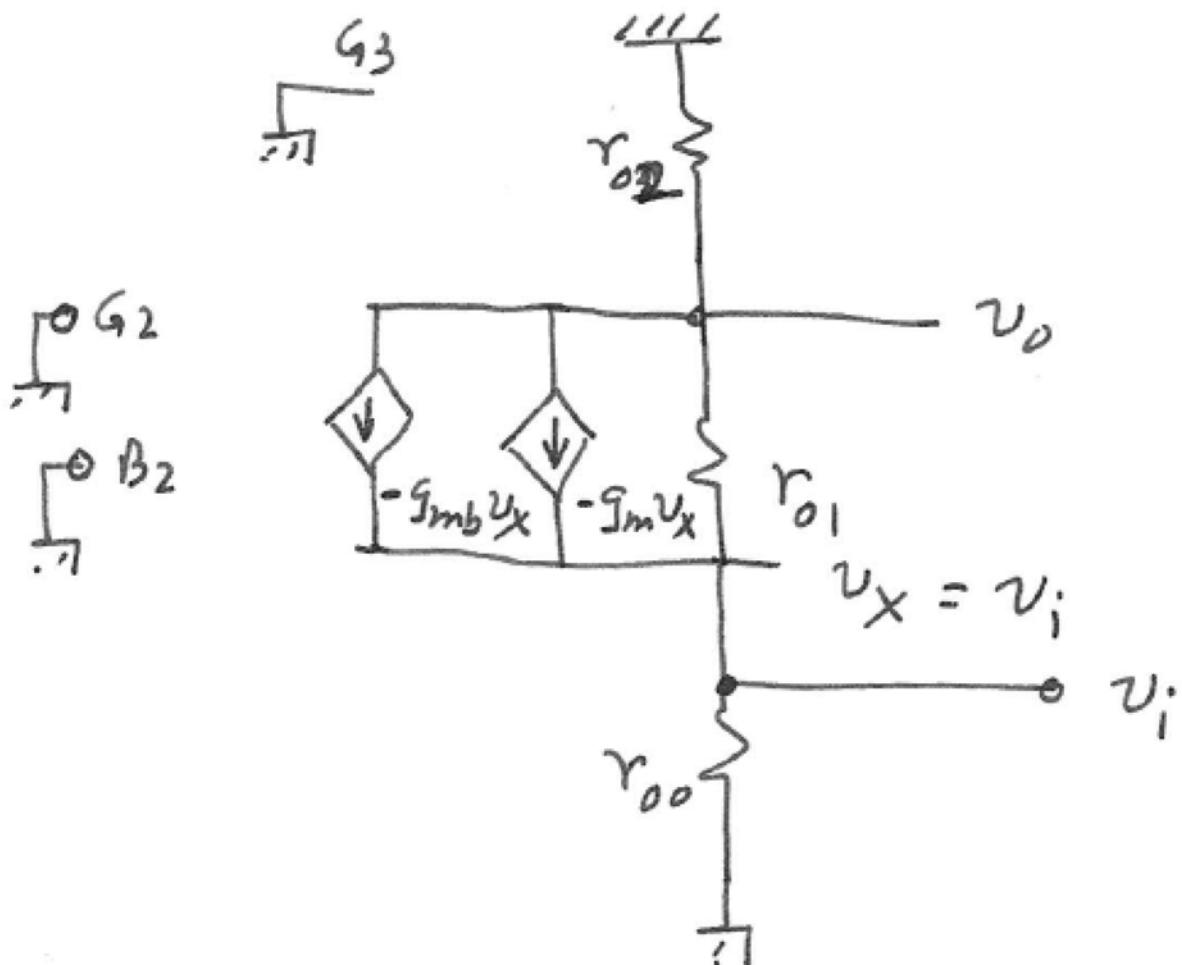
VAx := 50
> rox:=VAx/ID;
rox := 250000.0000
> RD:=5000;
RD := 5000
> Vgain:=-gm*(rox*RD)/(rox+RD);
Vgain := -9.803921569

```

The small signal voltage gain is -9.8 V/V

Q.8

The ac equivalent circuit is (hand drawing):



KCL at v_o node (note $v_x = v_i$, the input signal voltage)

$$\frac{v_o}{r_{o2}} - (g_m + g_{mb})v_i + \frac{v_o - v_i}{r_{o1}} = 0$$

> $y := \frac{v_o}{ro2} - (gm + gmb) vi + \frac{vo - vi}{rol}$

> $\text{solve}(y, vo);$

$$vi (rol gm + rol gmb + 1) ro2$$

> $gain := (rol * gm + rol * gmb + 1) * ro2 / (rol + ro2);$

$$gain := \frac{(rol gm + rol gmb + 1) ro2}{rol + ro2}$$

The voltage gain is: $\frac{v_o}{v_i} = \frac{1.2g_m r_{o1} + 1}{r_{o1}/r_{o2} + 1}$ (since, $g_{mb} = 0.2g_m$)