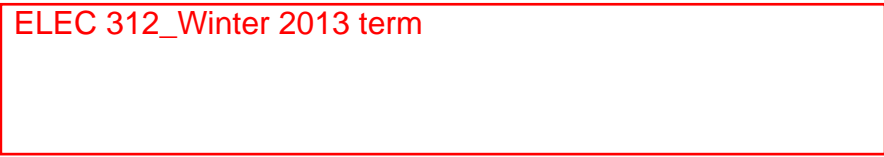




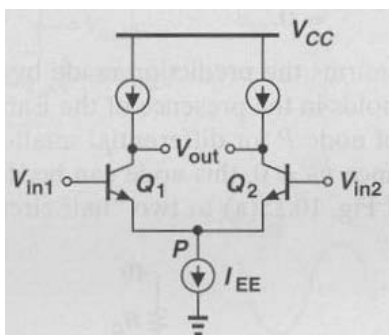
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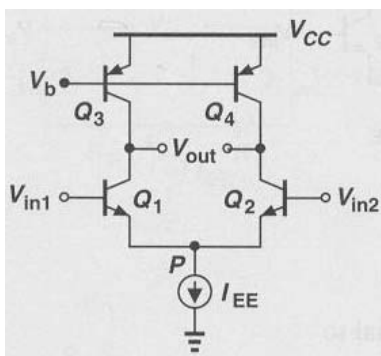
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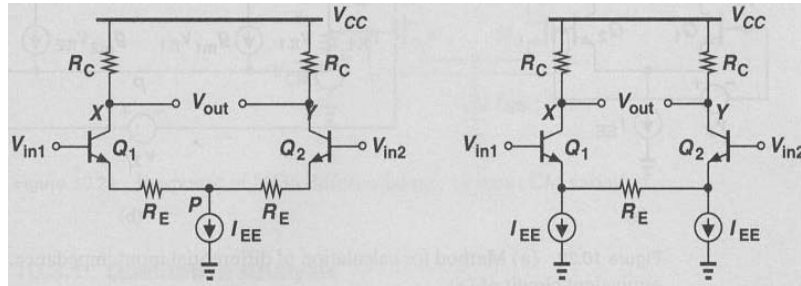
1. Find an expression for the differential gain of the following circuit, where ideal current sources are used as loads to maximize the gain. V_{in1} , V_{in2} may be assumed to be balanced differential signals.



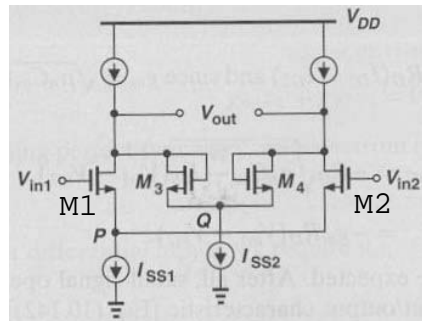
2. The following figure illustrates an implementation of a differential amplifier with active load using complementary BJT devices. Calculate the differential voltage gain $V_{out}/(V_{in1} - V_{in2})$.



3. Determine the gain of the emitter degenerated differential pairs shown in the following figure. Assume $V_A = \infty$.



4. Assuming $\lambda = 0$, compute the voltage gain of the following circuit. I_{SS2} is used to bias the transistors M_3, M_4 . Consider all I-sources as ideal.



5. Consider the basic current mirror in Figure 5 built with the NMOS-transistors M1, M2. For both the transistors $L=0.5$ microns, $W/L=8$, $V_{THN}=0.5V$, $\mu_n C_{oxn} = 300 \mu A/V^2$ Further, $V_{DD}=1.5V$, and $I_{REF}=100 \mu A$.

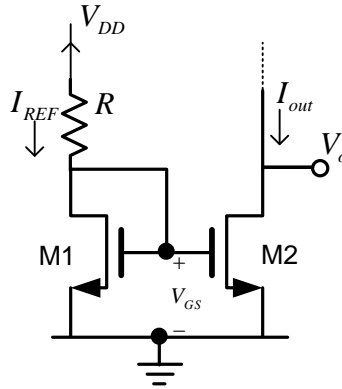


Figure 5

- (i) Design R for $I_{out} = 100 \mu A$.
- (ii) What is the lowest value that could be allowed for V_o for proper operation of the system?
- (iii) If the transistors have an *Early voltage* of $20V$, what is the output resistance of the current source (i.e., at the location of I_{out})?
- (iv) How much (as a %) will I_{out} change if V_o changes by $\pm 0.5V$?

6. Consider the basic current mirror (Figure 6) implemented using *npn*-BJT devices (Q1,Q2). Derive an expression for the current transfer ratio I_{out}/I_{in} . If β of the transistors has a minimum value of 50 , what will be the largest current transfer ratio?

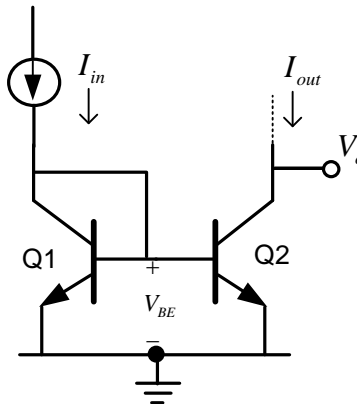
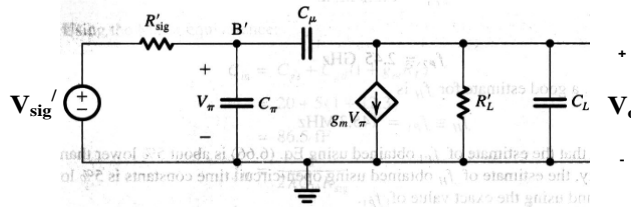


Figure 6

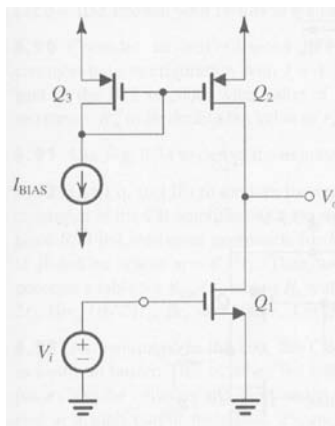
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1. A common-emitter amplifier that can be represented by the following equivalent circuit, has $C_\pi = 10$ pF, $C_\mu = 0.5$ pF, $C_L = 2$ pF, $g_m = 20$ mA/V, $\beta = 100$, $r_x = 200$ Ω , $R_L' = 5$ k Ω and $R_{sig} = 1$ k Ω . Find (i) the mid band gain A_M , (ii) the frequency of the zero f_Z , and (iii) the approximate values of the pole frequencies f_{P1} and f_{P2} . Hence estimate the 3-dB frequency f_H . Note that R'_{sig} is the equivalent Thevenin resistance looking towards the signal source and includes the effects of R_{sig} , r_x and r_π . For approximate estimates, you may use OCTC method.



2. Analyze the high-frequency response of the CMOS amplifier shown below. The dc bias current is 100 μ A. For Q_1 , $\mu_n C_{ox} = 90$ μ A/V², $V_A = 12.8$ V, $W/L = 100$ μ m/1.6 μ m, $C_{gs} = 0.2$ pF, $C_{gd} = 0.015$ pF. For Q_2 , $C_{gd} = 0.015$ pF, $C_{gs} = 36$ fF and $|V_A| = 19.2$ V. Assume that the resistance of the input signal generator is negligibly small. Also, for simplicity assume that the signal voltage at the gate of Q_2 is zero. Find the low-frequency (i.e., at DC) gain, the frequency of the pole, and the frequency of the zero. You may use nodal analysis.

Note: fF=10⁻¹⁵ F, pF=10⁻¹² F.

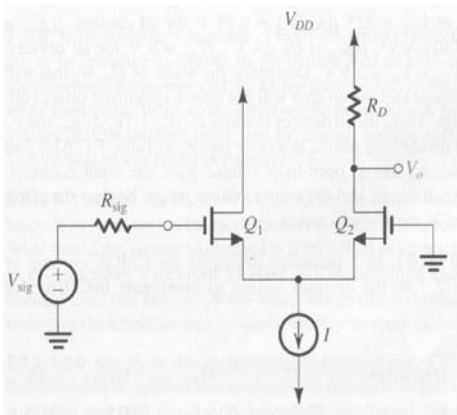


3. A CG amplifier is specified to have $C_{gs} = 2$ pF, $C_{gd} = 0.1$ pF, $C_L = 2$ pF, $g_m = 5$ mA/V, $\chi = 0.2$, $R_{sig} = 1$ k Ω and $R_L' = 20$ k Ω . Neglecting the effects of r_o , find the low-frequency gain v_o/v_{sig} , the frequencies of the poles f_{P1} and f_{P2} and hence an estimate of the 3-dB frequency f_H . For a CG amplifier you can use $g_{mb} = \chi g_m$. Use ac equivalent circuit.

4. (a) Consider a CS amplifier having $C_{gd} = 0.2$ pF, $R_{sig} = R_L = 20$ k Ω , $g_m = 5$ mA/V, $C_{gs} = 2$ pF, C_L (including C_{db}) = 1 pF, and $r_o = 20$ k Ω . Find (i) the low-frequency gain A_M , and (ii) estimate f_H using open-circuit time constants.

Hence determine the gain-bandwidth (GBW=mid-freq. gain *times* f_H).

5. Consider the following circuit for the case: $I = 200$ μ A and $V_{OV} = 0.25$ V, $R_{sig} = 200$ k Ω , $R_D = 50$ k Ω , $C_{gs} = C_{gd} = 1$ Pf (for both transistors). Find the dc (i.e., low-frequency) gain, the high-frequency poles, and an estimate of f_H . (hint: need to find g_m from I and V_{OV} data!).





6: (a) Consider a CS stage having $C_{gd} = 0.2 \text{ pF}$, $R_{sig} = 20 \text{ k}\Omega$, $g_m = 5 \text{ mA/V}$, $C_{gs} = 2 \text{ pF}$, and $r_o = 20 \text{ k}\Omega$.

(b) A CG stage is connected in totem-pole configuration with the CS transistor in (a) to create a cascode amplifier. The ac parameters of this stage are identical with those of the CS stage. Regarding the body-effect in the CG stage assume $\chi = 0.2$. Further $R_L = 20 \text{ k}\Omega$, and is shunted by a load capacitance $C_L = 1 \text{ pF}$.

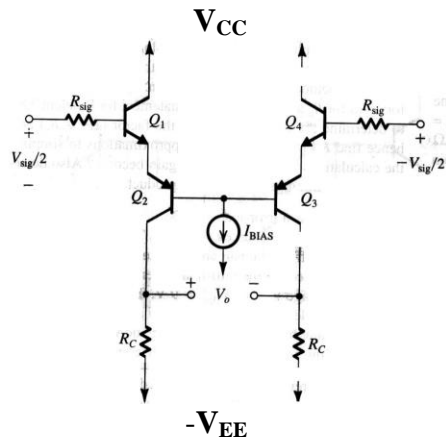
Show a schematic diagram of the system using NMOS transistors. Show the ac equivalent circuit.

Find (i) the low-frequency gain A_M , and (ii) estimate the gain-bandwidth of the system. You may use OCTC method to determine the dominant high frequency pole f_H of the system.

7: For the following circuit, let the bias be such that each transistor is operating at $100\text{-}\mu\text{A}$ collector current. Let the BJTs have $h_{fe} = 200$, $f_T = 600 \text{ MHz}$, and $C_\mu = 0.2 \text{ pF}$, and neglect r_o and r_x . Also, $R_{sig} = R_C = 50 \text{ k}\Omega$.

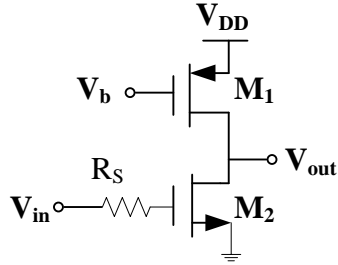
Show the ac equivalent circuit.

Find (i) the low-frequency gain, (ii) the high-frequency poles, and (iii) an estimate of the dominant high frequency pole f_H of the system. Now find the GBW (gain-bandwidth) of the system. You may use half-circuit technique.





8: In the following circuit assume both transistors operate in saturation and $\lambda \neq 0$. For each transistor you can assume the parasitic capacitances as $C_{gsi}, C_{gdi}, (i=1,2)$.



Draw the *ac* equivalent circuit, analyze and derive the expression for the dominant pole frequency.

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1. A series-series feedback circuit represented by Fig.1, transconductance amplifier operates with $V_s = 100$ mV, $V_f = 95$ mV, and $I_o = 10$ mA. What are the corresponding values of A and β ? Include the correct units for each.

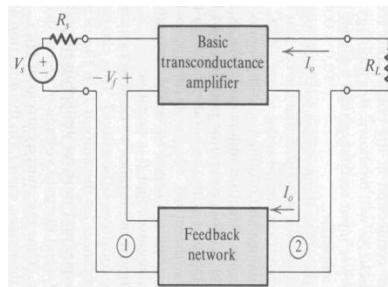


Figure 1:

2. For an amplifier connected in a negative feedback loop in which the output voltage is sampled (i.e., a shunt connection), measurement of the output resistance before and after the loop is connected shows a change by a factor of 80. Is the resistance with feedback higher or lower? What is the value of the loop gain $A\beta$? If R_{of} is 100Ω , what is R_o without feedback?
3. The shunt-shunt feedback amplifier in the Figure 3 has $I = 1$ mA and $V_{GS} = 0.8$ V. The MOSFET has $V_t = 0.6$ V and $V_A = 30$ V. For $R_s = 10$ k Ω , $R_1 = 1$ M Ω , and $R_2 = 4.7$ M Ω , find the voltage gain v_o/v_s , the input resistance R_{in} and the output resistance R_{out} . You need to figure out the *ac* parameters for the MOS device.

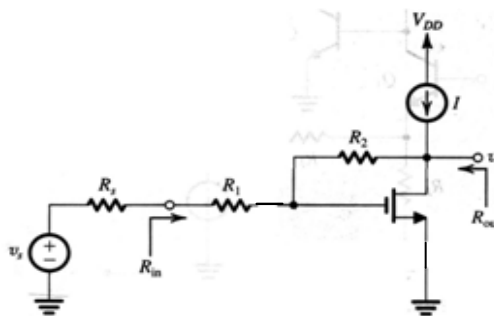


Figure 3:

4. An op amp having a low-frequency gain of 10^3 and a single-pole transfer function with -3dB frequency of 10^4 rad/s is connected in a negative feedback loop via a feedback network having a transmission $\beta(s)$ given by $\beta(s) = \frac{\beta_o}{(1 + s/10^4)^2}$. Find the value of β_o above which the closed-loop amplifier becomes unstable.



5. A DC amplifier has an open-loop gain of 1000 and two poles, a dominant one at 1 kHz and a high-frequency one whose location can be controlled. It is required to connect this amplifier in a negative feedback loop that provides a dc closed-loop gain of 100 and a maximally flat response. The transfer function of the amplifier can be modeled as:

$$A(s) = \frac{1000}{(1 + s/\omega_1)(1 + s/\omega_2)}$$

In the above ω_1 is the dominant pole frequency. It is required that under feedback, the amplifier will have a maximally flat response according to the model

$$A_f(s) = \frac{1000\omega_1\omega_2}{s^2 + (\omega_p/Q_p)s + \omega_p^2}, \text{ with } Q_p=0.707.$$

Calculate the required ω_2 .

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Q.1: Show the design of a class-A power amplifier using BJT devices, employed to deliver 10 W of ac signal power to a load of 10 ohms. Find the V_{CC} required, the I_Q required and design the active circuit to provide the required I_Q . Comment on the heat dissipation limits of the BJT devices used in your design.

Q.2: Design an idealized class-B output stage as shown in Figure 2, to deliver an average power of 25 W to an 8 Ω speaker. The peak output voltage must be no larger than 80% of supply voltages V_{CC} . Assume that the input signal is sinusoidal.

Determine: (i) the required value of V_{CC} , (ii) the peak current in each transistor, (iii) the average power dissipated in each transistor, and (iv) the power conversion efficiency.

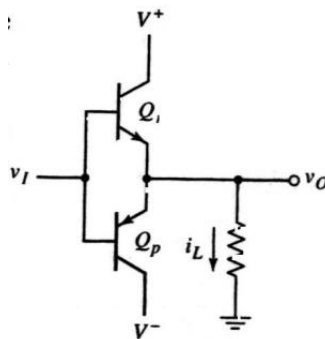


Figure 2: Basic class-B output stage

Q.3: Determine the required biasing in a MOSFET class-AB output stage. The circuit is shown in Figure 3. The parameters are $V_{DD} = 10$ V and $R_L = 20$ Ω . The transistors are matched, and the parameters are $K = 0.20$ A/V² and $|V_T| = 1$ V. The quiescent drain current is to be 20% of the load current when $v_o = 5$ V.

The I-V equation for either transistor is: $I_D = K(V_{GS} - |V_T|)^2$

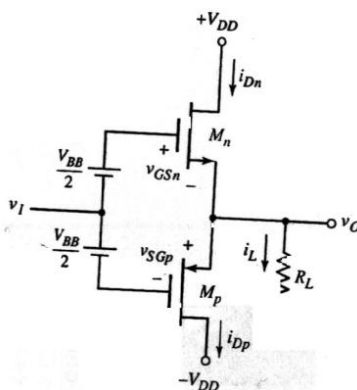


Figure 3: MOSFET class-AB output stage