

$$D(s) = s^2 + \frac{s}{R_1 C_1} \cdot \frac{1+R_6/R_5}{1+R_4/R_3} + \frac{R_6/R_5}{R_1 R_2 C_1 C_2}$$

$$\rightarrow s^2 + s \left( \frac{\omega_p}{Q_p} \right) + \omega_p^2$$

Thus:

$$\frac{R_6/R_5}{R_1 R_2 C_1 C_2} = (10^5)^2$$

$$\frac{1}{R_1 C_1} \cdot \frac{1+R_6/R_5}{1+R_4/R_3} = \frac{\omega_p}{Q_p} = BW = 10^3$$

Let  $R_1 = R_2 = R$ ;  $C_1 = C_2 = C$ . } 8 elements  
 $R_3, R_4, R_5, R_6, R, C$  to be determined. } 2 conditions

Then

$$\frac{1}{RC} \cdot \frac{1+R_6/R_5}{1+R_4/R_3} = 10^3$$

$$\left( \frac{1}{RC} \right)^2 \frac{R_6}{R_5} = (10^5)^2 = 10^{10}$$

TBD  $\rightarrow$  to be determined.

Let  $R_5 = R_6 = R_x$  (so  $R_3, R_4, R_x, R, C$  TBD)

$$\frac{1}{RC} = 10^5$$

Let  $R = 1$ ,  $C = 10^{-5}$  (so  $R_3, R_4, R_x$  TBD).

Going back:

$$\frac{1}{R_1 C_1} \cdot \frac{1+R_6/R_5}{1+R_4/R_3} = 10^3$$

$$\text{ie. } 10^5 \cdot \frac{1+1}{1+R_4/R_3} = 10^3$$

$$1 + \frac{R_4}{R_3} = 200 \quad ; \quad \frac{R_4}{R_3} = 199.$$

Q1:

Let  $R_3 = 1, R_4 = 199; R_x = 1.$

So  $R_1 = R_2 = 1; R_3 = 1, R_4 = 199; R_5 = R_6 = 1$   
 $C_1 = C_2 = 10^5$

is the FIRST cut design.

Verify:  $\omega_p$  &  $\omega_p$  values.

Practical set of values? Let  $R_1 = R_2 = R_3 = R_5 = R_6 = 100 \Omega$

Then  $R_4 = 19900 = 19.9 \text{ k}\Omega$

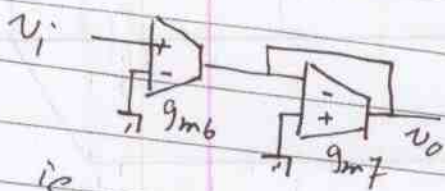
$C_1 = C_2 = 10^{-7} = 0.1 \mu\text{F}.$

x

Q2: Using the general structure (5-OTA) and with  $V_A, V_B = 0$ , we'll have 3 OTAs only.

$$\frac{v_o}{v_i} = \frac{s^2 C_1 C_2}{s^2 C_1 C_2 + s G_1 G_2 + G_1 G_2}$$

At  $\omega \rightarrow \infty$ , gain = 1. To get  $H_0 = 3$ , we can use an OTA-based gain-stage ic.



$$\frac{v_o}{v_i} = 3 = \frac{g_{m6}}{g_{m7}}; \therefore g_{m6} = 3g_{m7}$$

ie  $g_{m6} = 3g_{m7}.$

Further:

$$\frac{g_{m1} g_{m2}}{C_1 C_2} = (2000)^2; \frac{g_{m3}}{C_2} = \frac{2000}{5} = 400$$

Q2: We could choose  $C_1 = C_2 = 0.01 \mu F$ .

$g_{m1} = g_{m2} = g_m$  ;  $\frac{g_m}{0.01 \mu F} = 2000$  ;  $g_m = 20 \mu S$

$Q_p = 5 = \frac{10^{-8}}{g_{m3}} \cdot 2000$

$\frac{2000}{5} = \frac{g_{m3}}{C}$  ;  $g_{m3} = \frac{2000 \times 10^{-8}}{5} = 4 \mu S$

Let  $g_{m7} = 4 \mu S$ ,  $g_{m6} = 12 \mu S$

Total  $g_m$ :  $= \sum g_m = 60 \mu S$

For the BJT stage  $g_m|Q = \frac{I_{DC}}{V_T}$  ;  $V_T = 25 mV$ .

So  $\{ I_{DC} = 25 \times 10^{-3} \times 60 \times 10^{-6} = 1.5 \times 10^{-6}$

$I_{Bias} = 2 \sum I_{DC} = \dots$

$P_{DC} = (5 - (-5)) \times 3 \mu A = 30 \mu W$

Q3: Using given data find the order  
(hint)  $n=5$  for the normalized LPF (CHEB filter)

For CHEB filter of order 5 ( $A_p = 0.5 dB$ )

$H_N(s)|_{LP} = \frac{1}{2^4 \cdot \epsilon \cdot \sqrt{s^5 + 1.1725s^4 + 1.9374s^3 + 1.3096s^2 + 0.7525s + 0.1789}}$

$\epsilon = 0.3403$

Use frequency scaling  $s \rightarrow \frac{\omega_{CH}}{3} s$  ;  $\omega_{CH} = 2\pi \times 15 \times 10^3$   
Apply BLT  $s \rightarrow 2 \times 150 \times 10^3 \left( \frac{s-1}{2} \right) / \left( \frac{s+1}{2} \right)$  to get  $H(z)$

Q4:

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(hint)

$$V_1^{(1)} = V_i^{(1)} \frac{1}{1-z^{-1}} \cdot \frac{-G}{D} + V_i^{(2)} \frac{z^{-1/2}}{1-z^{-1}} \cdot \frac{H}{D} + V_2^{(1)} \frac{1}{1-z^{-1}} \cdot \frac{-C}{D}$$

$$\therefore V_i^{(2)} = z^{-1/2} V_i^{(1)}$$

$$V_1^{(1)} = V_i^{(1)} \left[ \frac{-\frac{G}{D} + \frac{z^{-1/2} H}{D}}{1-z^{-1}} \right] + V_2^{(1)} \cdot \frac{-C/D}{1-z^{-1}}$$

Similarly:

$$V_2^{(1)} = V_i^{(2)} \cdot \frac{z^{-1/2}}{1-z^{-1}} \cdot \frac{A}{B} - V_2^{(1)} \cdot \frac{1}{1-z^{-1}} \cdot \left( -\frac{F}{B} \right)$$

$$V_2^{(1)} \left[ 1 + \frac{F/B}{1-z^{-1}} \right] = V_i^{(1)} \frac{z^{-1/2}}{1-z^{-1}} \left( \frac{A}{B} \right)$$

$$\text{Then: } V_2^{(1)} \left[ 1 + \frac{F/B}{1-z^{-1}} \right] = \frac{A}{B} \cdot \frac{z^{-1/2}}{1-z^{-1}} \times [ \quad ]$$

$$\left[ V_i^{(1)} \frac{-G/D + z^{-1/2} H/D}{1-z^{-1}} + V_2^{(1)} \frac{-C/D}{1-z^{-1}} \right]$$

Simplify to get

$$\frac{V_2^{(1)}}{V_i^{(1)}} = \dots$$

Q1:  $\omega_p = 10^5$ ,  $BW = 10^3$ ,  $Q_p = 10^5 / 10^3 = 100$ .

A BPF has

$$H(s) = H_0 \frac{(\omega_p / Q_p) s}{s^2 + (\omega_p / Q_p) s + \omega_p^2}$$

So we must have in the

given  $\frac{v_o}{v_i}$ ,  $R_8 / R_6 \rightarrow 0$  i.e.  $R_8 \rightarrow \infty$ , open.

$R_8 / (R_3 R_5 R_7 C_1 C_2) \rightarrow 0$  i.e.  $R_8 \rightarrow 0$ , short, or  
one of  $R_3, R_5, R_7 \rightarrow \infty$  i.e. open.

$$\frac{1}{R_1 C_1} \left( \frac{R_8}{R_6} - \frac{R_1 R_8}{R_4 R_7} \right) = H_0 \cdot (\omega_p / Q_p)$$

So  $R_8 \rightarrow 0$  is not possible; we take  $R_6 \rightarrow \infty$ .

Then  $\frac{v_o}{v_i} = - \frac{\frac{1}{R_1 C_1} \cdot \frac{R_1 R_8}{R_4 R_7} s}{s^2 + \frac{1}{R_1 C_1} s + \frac{R_8}{R_2 R_3 C_1 C_2 R_7}}$

In the choice  $R_3, R_5, R_7$  one of them  $\rightarrow \infty$ , only  
choice is  $R_5 \rightarrow \infty$   $\because R_3, R_7$  decides  $\omega_p$ .

Now:  $\frac{R_8}{R_2 R_3 C_1 C_2 R_7} = \omega_p^2 = 10^{10}$   
 $\frac{1}{R_1 C_1} = \frac{\omega_p}{Q_p} = 10^3$   $H_0 \rightarrow$  not specific

Q1:

To determine  $\omega_p, Q_p$ : we have $C_1, C_2, R_2, R_3, R_7, R_8, R_4, R_1 \rightarrow 8$  elementsLet  $C_1 = C_2 = C$ ;  $R_2 = R_3 = R_7 = R_8 = R$ Then  $C, R, R_4, R_1$  need be found out.

$$\frac{1}{CR} = 10^5 \quad ; \quad \frac{1}{R_1 C} = 10^3$$

$$\text{Let } R=1; C=10^{-5} \quad ; \quad \frac{1}{R_1} = 10^3 \times 10^{-5} = 10^{-2}$$

$$R_1 = 100$$

$$\text{So } R_1 = 100, R_2 = R_3 = R_7 = R_8 = R = 1$$

$$C_1 = C_2 = C = 10^{-5} \quad ; \quad R_4 \rightarrow \text{free choice} = 1 (\text{say})$$

First cut design values

$$R_1 = 100; R_2 = R_3 = R_4 = R_7 = R_8 = 1.$$

$$C_1 = C_2 = 10^{-5}$$

Let  $R = 100 \Omega$ ; Then

$$R_1 = 10^4 = 10 \text{ k}\Omega$$

$$R_2 = R_3 = R_4 = R_7 = R_8 = 100 \Omega$$

$$C_1 = C_2 = 10^{-7} = 0.1 \text{ }\mu\text{F.}$$

 $R_6, R_5$  open  $\rightarrow$  i.e. absent in the circuit.

- Q2, Q3  $\rightarrow$  See the ELEC 441 (UG class) soln / hint

Q4: Pre-warp:  $\hat{\omega}_c = 2f_s \cdot \tan\left(\frac{\omega_c}{2f_s}\right)$

$$= 2f_s \cdot \tan\left(\frac{2\pi \times 3.4 \times 10^3}{2 \times 64 \times 10^3}\right)$$

$$= 2f_s \cdot 0.168 = a \times 0.168$$

$$f_2 = (2f_s)^2 + \left(\frac{\hat{\omega}_c}{Q_p}\right) \cdot 2f_s + \hat{\omega}_c^2$$

$$a = 2f_s$$

$$= (2f_s)^2 \left[ 1 + \frac{0.168}{1.3} + 0.168^2 \right]$$

Then from the Table:

$$H(z) = h_D \frac{1 + 2\bar{z}^{-1} + \bar{z}^{-2}}{1 - a_{1D}\bar{z}^{-1} + a_{2D}\bar{z}^{-2}}$$

DC gain of 0 dB  $\rightarrow 1 = h_D \cdot \frac{2+2}{1-a_{1D}+a_{2D}}$

where  $a_{1D} = \frac{2[(2f_s)^2 - (2f_s)^2(0.168)^2]}{(2f_s)^2 \left[ 1 + \frac{0.168}{1.3} + 0.168^2 \right]}$   $= \frac{2(1-0.168^2)}{1 + \frac{0.168}{1.3} + 0.168^2}$

$$a_{2D} = \frac{(2f_s)^2 \left[ 1 - \frac{0.168}{1.3} + 0.168^2 \right]}{(2f_s)^2} = \frac{\left[ 1 - \frac{0.168}{1.3} + 0.168^2 \right]}{1}$$

Thus,  $a_{1D} = 1.679$

$$a_{2D} = 0.777$$

So  $h_D = \frac{1 - 1.679 + 0.777}{4} = 0.0245$

Q4.

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Given  $V_2/V_1$  can be re-written as:(set  $L, I, J, E = 0$ , these are not in the given schematic)

$$= - \frac{A(z_G - H) + D(z-1)^2 K}{A z C + D F z(z-1) + D B (z-1)^2}$$

$$= -DK \cdot \left[ 1 + \frac{A_G - 2DK}{DK} z^{-1} + \frac{DK - AH}{DK} z^{-2} \right]$$

$$\frac{(DF + DB)}{(DF + DB)} \left[ 1 - \frac{2DB + DF - AC}{DF + DB} z^{-1} + \frac{DB}{DF + DB} z^{-2} \right]$$

$$= - \frac{K}{F+B} \cdot \frac{1 + \left( \frac{A_G}{DK} - 2 \right) z^{-1} + \left( 1 - \frac{AH}{DK} \right) z^{-2}}{1 - \frac{2DB + DF - AC}{DF + DB} z^{-1} + \frac{B}{F+B} z^{-2}}$$

Design for

$$\frac{K}{F+B} = h_d = 0.0245$$

$$\frac{A_G}{DK} - 2 = 2$$

$$1 - \frac{AH}{DK} = 1$$

$$\frac{2DB + DF - AC}{DF + DB} = a_{1D} = 1.679$$

$$\frac{B}{F+B} = a_{2D} = 0.777$$

$$\text{Let } B = D = 1$$

5 equations

A, B, C, D, F, G, K, H  
B capacitors.



Q4:

$$\frac{1}{1+F} = 0.777 \quad ; \quad F = \frac{1}{.777} - 1 = 0.287$$

$$\text{Then } \frac{2 + .287 - AC}{0.287 + .287} = 1.679$$

$$\text{So, } AC = 1.323$$

$$\frac{K}{1.287} = .0245 \quad , \quad K = 0.032$$

$$AG = 4 \times 1 \times .032 = 0.128$$

$$1 - \frac{AH}{DK} = 1 \quad \text{can be satisfied with } H=0$$

(i.e. H cap. is absent)

$$\text{So } B = D = 1 \quad ; \quad F = 0.287 \quad , \quad K = 0.032$$

A, C, G remaining.

If we set  $A=1$ ,  $C=1.323$ ,  $G=.128$   
 completes the first cut design. Thus

$$A = B = D = 1 \quad ; \quad C = 1.323, \quad F = 0.287, \quad G = 0.128, \\ K = 0.032$$

Scale by the lowest valued cap i.e.  $k = .032$

$$K = 1 \quad , \quad A = B = D = 31.25$$

$$C = 41.34 \quad , \quad F = 8.97 \quad , \quad G = 4$$

If  $C \rightarrow$  largest cap must be  $< 20 \text{ PF}$ , scale  
 again by 2.067

$$\text{Then } K = 0.483 \text{ PF}, \quad A = B = D = 15.12 \text{ PF}, \quad C = 20 \text{ PF} \\ F = 4.34 \text{ PF} \quad ; \quad G = 1.94 \text{ PF}.$$