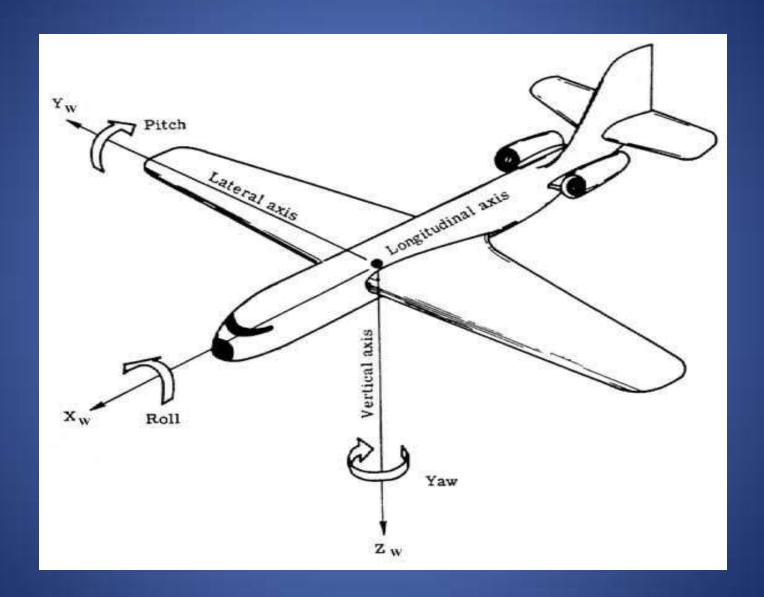
Modeling and Simulation of a Three Degree of Freedom Longitudinal Aero plane System



Figure 1: Boeing 777 and example of a two engine business jet



Nonlinear dynamic equations of motion for the longitudinal direction of aircraft

$$\begin{split} \dot{u} &= X/m - g\sin\theta + rv - qw \\ \dot{w} &= Z/m + g\cos\phi\cos\theta + qu - pv \\ \dot{x}_1 &= \left(\cos\theta\cos\psi\right)u + \left(-\cos\phi\sin\psi + \sin\phi\sin\theta\cos\psi\right)v \\ &+ \left(-\sin\phi\sin\psi + \cos\phi\sin\theta\cos\psi\right)w \\ \dot{z}_1 &= \left(-\sin\theta\right)u + \left(\sin\phi\cos\theta\right)v + \left(\cos\phi\cos\theta\right)w \\ \dot{q} &= \left[M - \left(I_{xx} - I_{zz}\right)pr - I_{xz}\left(p^2 - r^2\right)\right] \div I_{yy} \\ \dot{\theta} &= q\cos\phi - r\sin\phi \end{split}$$

Nonlinear dynamic equations of motion for the longitudinal direction of aircraft

- The longitudinal equations of motions are considered for the aircraft model with the following assumptions
- V=0
- Y=0
- p=0
- r=0
- Ø=0
- Ψ=0
- lxx=0
- |zz=0
- |xz=0

Nonlinear dynamic equations of motion for the longitudinal direction of aircraft

The equations after assumptions become:

$$\dot{u} = X/m - g\sin\theta - qw$$

$$\dot{w} = Z/m + g\cos\theta + qu$$

$$\dot{x}_1 = (\cos\theta)u + (\sin\theta)w$$

$$\dot{z}_1 = (-\sin\theta)u + (\cos\theta)w$$

$$\dot{q} = [M] \div I_{yy}$$

$$\dot{\theta} = q$$

The state vector can be shown as:

$$\begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \\ x_5 \\ x_6 \end{bmatrix} = X_{long} = \begin{bmatrix} u \\ w \\ z \\ z \\ \theta \end{bmatrix} = \begin{bmatrix} Axial & Velocity \\ Vertical & Velocity \\ Range \\ Altitude(-) \\ Pitch & Rate \\ Pitch & Angle \end{bmatrix}$$

The control input can be shown as:

$$\begin{bmatrix} u_1 \\ u_2 \\ u_3 \end{bmatrix} = U = \begin{bmatrix} \delta e \\ \delta t \\ \delta s \end{bmatrix} = \begin{bmatrix} Elevator & Command \\ Throttle, \% \\ stabulator \end{bmatrix}$$

Methods used to stabilize our model

- The Linear Quadratic Regulator (LQR)controller.
- ■The Proportional Integral Deferential (PID) controller.

Gain Calculations (Ziegler and Nichols Method)

P control
$$k_{p} = 0.5k_{pu}$$

PI control
$$\begin{cases} k_{p} = 0.45k_{pu} \\ k_{i} = 0.45k_{pu} / (0.83T_{u}) \end{cases}$$

PID control
$$\begin{cases} k_{p} = 0.6k_{pu} / (0.5T_{u}) \\ k_{d} = 0.6k_{pu} / (0.125T_{u}) \end{cases}$$

Gain Calculations (Transfer Functions)

$$\frac{X}{\&} = \frac{-7.501e - 006 \text{ s}^4 + 2757 \text{ s}^3 + 4246 \text{ s}^2 + 281.7 \text{ s} + 0.002549}{\text{s}^6 + 3.531 \text{ s}^5 + 17.14 \text{ s}^4 + 2.067 \text{ s}^3 + 0.1181 \text{ s}^2 + 1.071e - 006 \text{ s}}$$

$$\frac{Z}{\&} = \frac{-27.83 \text{ s}^3 - 83.67 \text{ s}^2 + 3947 \text{ s} + 67.91}{\text{s}^5 + 3.531 \text{ s}^4 + 17.14 \text{ s}^3 + 2.067 \text{ s}^2 + 0.1181 \text{ s} + 1.071e - 006}$$

$$\frac{\theta}{\&} = \frac{-18.81 \text{ s}^3 - 29.05 \text{ s}^2 - 0.6134 \text{ s} - 1.455e - 006}{\text{s}^5 + 3.531 \text{ s}^4 + 17.14 \text{ s}^3 + 2.067 \text{ s}^2 + 0.1181 \text{ s} + 1.071e - 006}$$

$$\frac{X}{\delta T} = \frac{4.674 \text{ s}^4 + 16.42 \text{ s}^3 + 78.44 \text{ s}^2 - 0.09065 \text{ s} + 6.172e - 017}{\text{s}^6 + 3.531 \text{ s}^5 + 17.14 \text{ s}^4 + 2.067 \text{ s}^3 + 0.1181 \text{ s}^2 + 1.071e - 006}$$

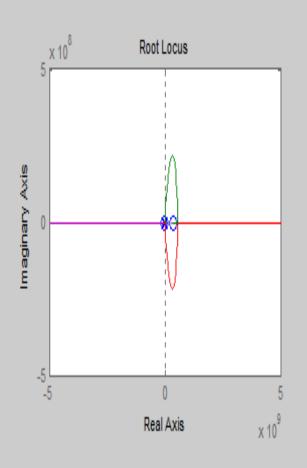
$$\frac{Z}{\delta T} = \frac{-0.1001 \text{ s}^3 - 0.7965 \text{ s}^2 - 2.517 \text{ s} - 8.463}{\text{s}^5 + 3.531 \text{ s}^4 + 17.14 \text{ s}^3 + 2.067 \text{ s}^2 + 0.1181 \text{ s} + 1.071e - 006}$$

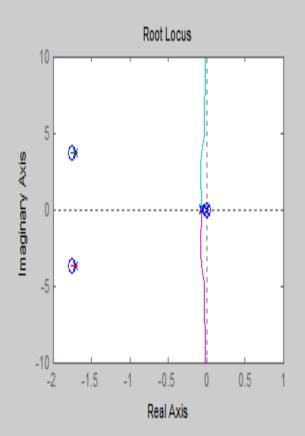
$$\frac{\theta}{\delta T} = \frac{0.009251 \text{ s}^2 + 0.05644 \text{ s} + 5.875e - 019}{\text{s}^5 + 3.531 \text{ s}^4 + 17.14 \text{ s}^3 + 2.067 \text{ s}^2 + 0.1181 \text{ s} + 1.071e - 006}$$

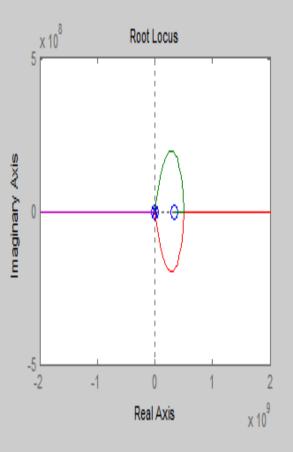
$$\frac{X}{\delta S} = \frac{-1.393e - 005 \text{ s}^4 + 4629 \text{ s}^3 + 7058 \text{ s}^2 + 468.3 \text{ s} + 0.004237}{\text{s}^6 + 3.531 \text{ s}^4 + 17.14 \text{ s}^4 + 2.067 \text{ s}^3 + 0.1181 \text{ s}^2 + 1.071e - 006 \text{ s}}$$

$$\frac{Z}{\delta S} = \frac{-52.06 \text{ s}^3 - 150.4 \text{ s}^2 + 6553 \text{ s} + 112.8}{\text{s}^5 + 3.531 \text{ s}^4 + 17.14 \text{ s}^3 + 2.067 \text{ s}^2 + 0.1181 \text{ s} + 1.071e - 006}$$

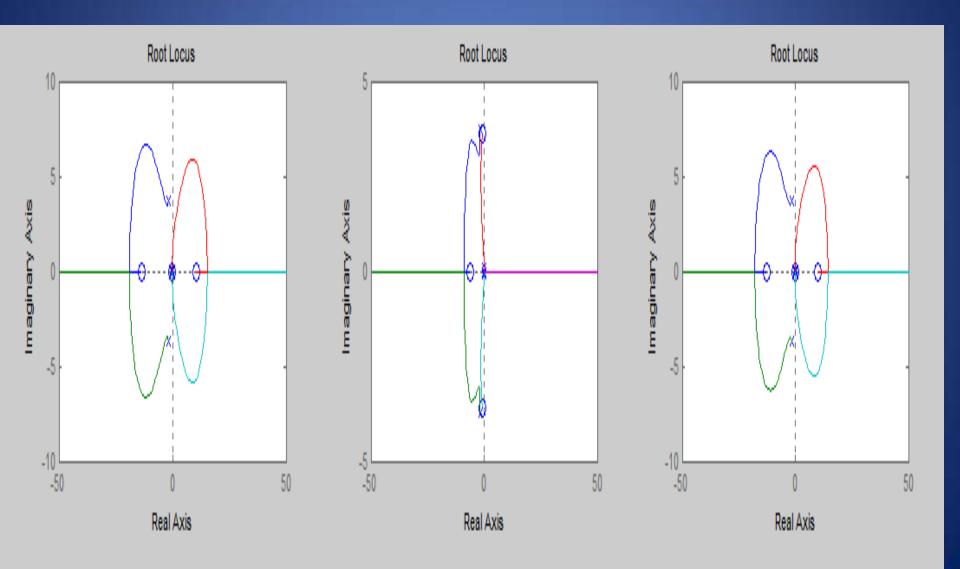
Root Locus Plots



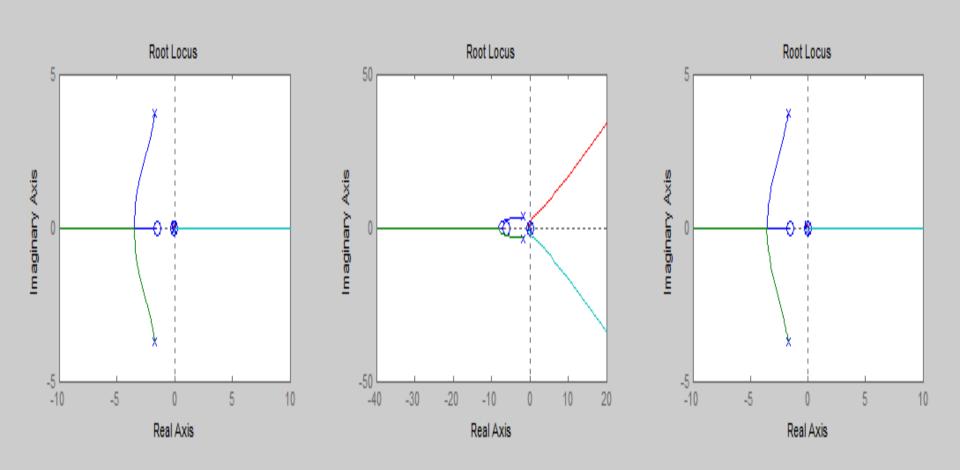




Root Locus Plots



Root Locus Plots



$$J = \int_{to}^{t} e^2 + \lambda u^2 dt$$

J is the energy spent by the actuators in order to regulate the system towards equilibrium

$$e \approx (X - Xcomm)$$

The error vector (e) is defined as the difference between the actual state vector, and the commanded value

>λ is a Lagrange multiplier

If a different value of weighting is required on each of the elements of the error vector and input u:

- ■A square matrices, denoted here as Q, R (identity matrices), are used to ensure that J is non-negative for all values of e, and u, but is zero when X and Xcomm (no inputs) are equal.
- ■For each choice of Q, and R, minimization of J corresponds to a unique choice of x, using specific inputs.
- ■Essentially, the ratio between Q and R matrices represents the effort on actuators .

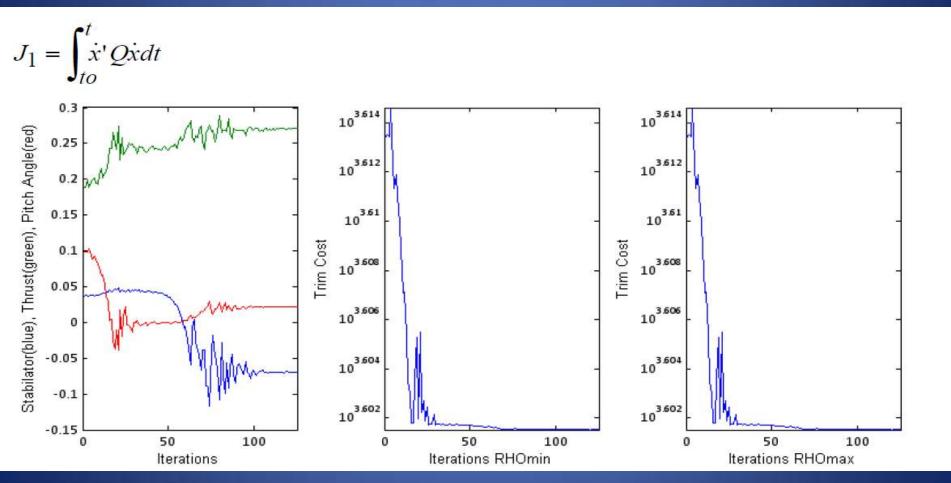
$$Q \gg R$$

The error is penalized, therefore the performances are maximized at the cost of an important effort on the actuators

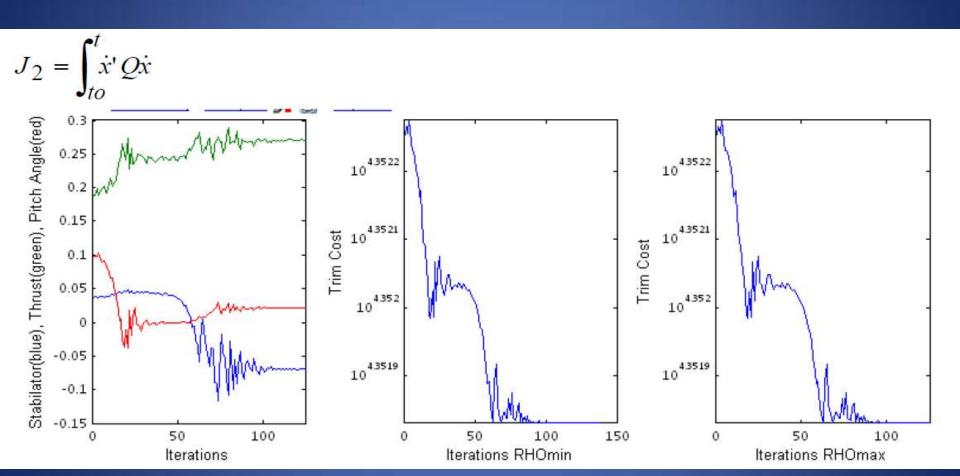
The control effort is penalized, therefore the energy used to compensate is reduced at the cost of lower performances.

$$Q = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & \dots & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}_{n \times n} R = \rho \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & \dots & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}_{r \times r}$$

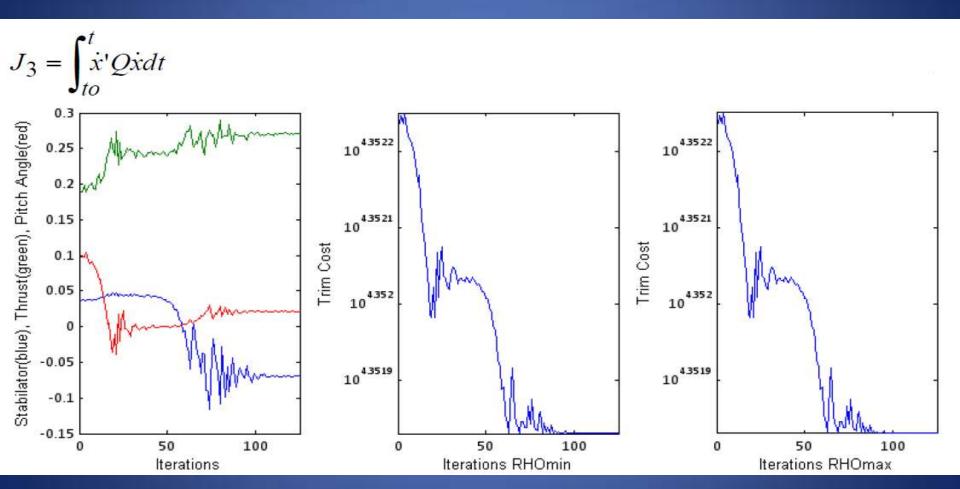
Cost Function Results



Highest performance, lowest cost, all state variables, and no inputs were used

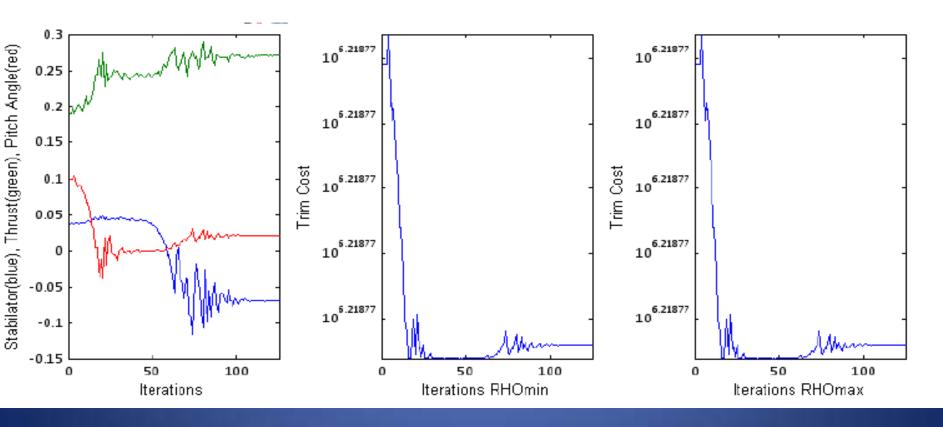


Lower performance, higher cost, all state variables, no inputs, but integration without dt

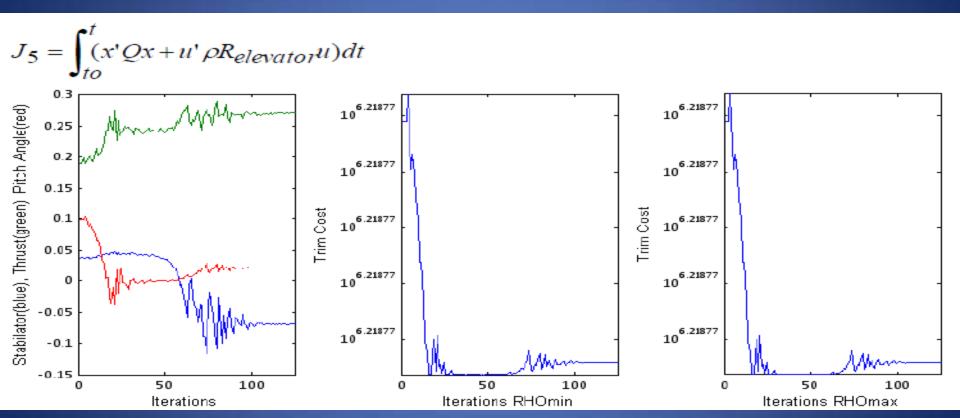


Lower performance, higher cost, not all state variables, no inputs, and specific state variables were used for compromise between cost and performance

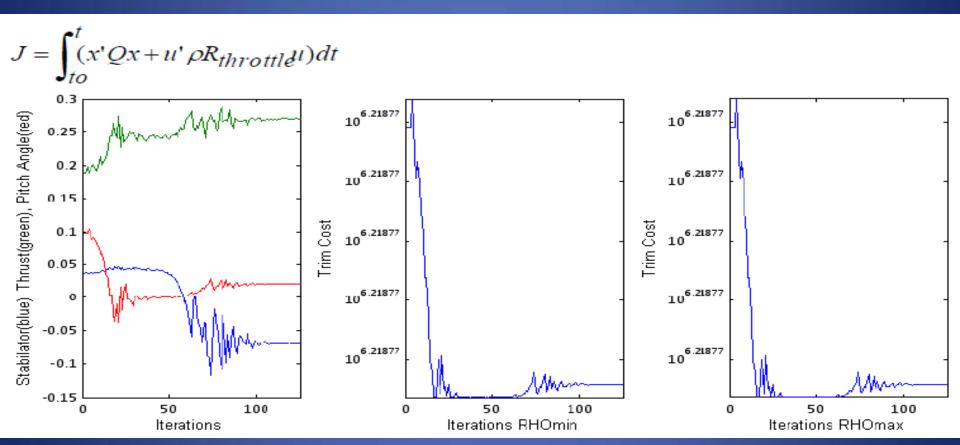
$$J = \int_{to}^{t} x' Qx dt$$



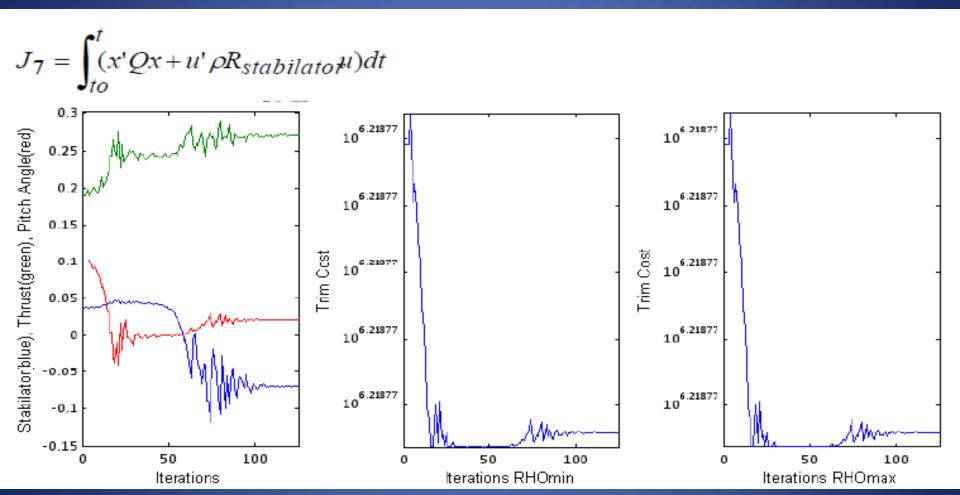
High performance, higher cost, all state variables, and no inputs were used for compromise between cost and performance



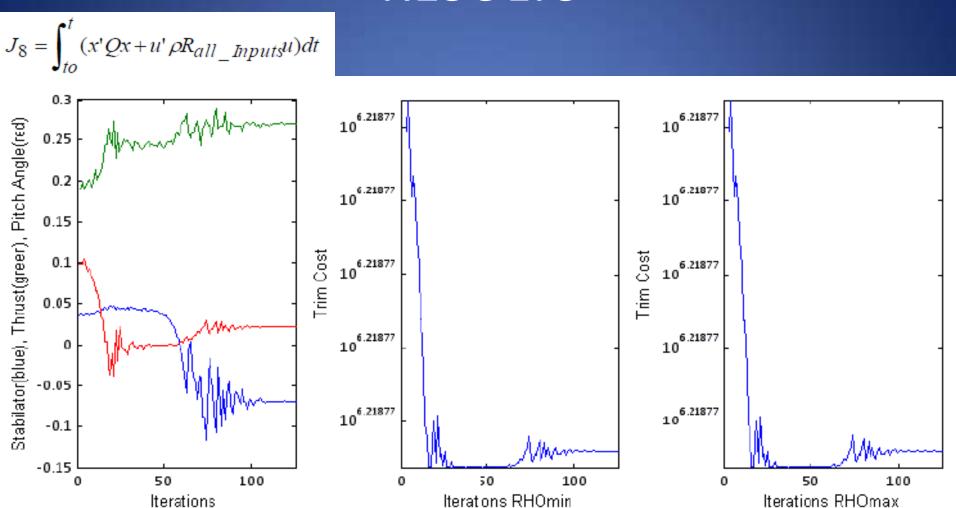
High performance, higher cost, all state variables, and elevator input is used only for reducing cost



High performance, higher cost, all state variables, and throttle input is used only for reducing cost



High performance, higher cost, all state variables, and stabilator input is used only for reducing cost



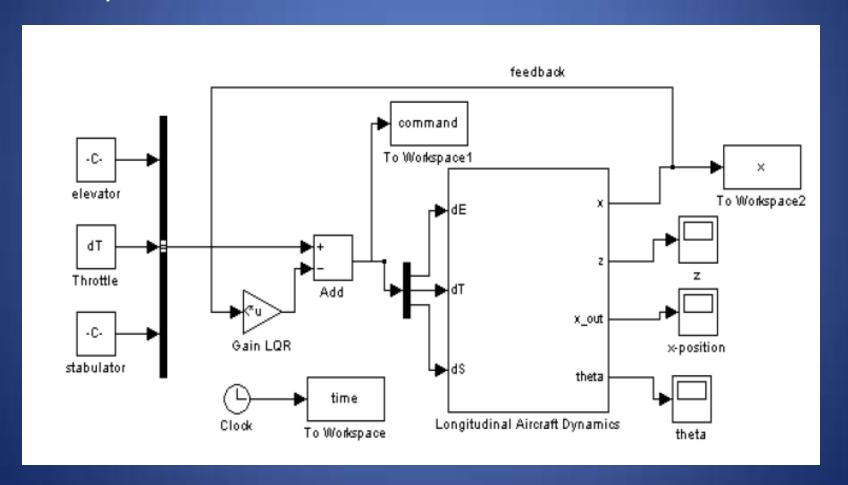
High performance, highest cost, all state variables, and all inputs were used (which is our real case)

Cost Function Conclusion

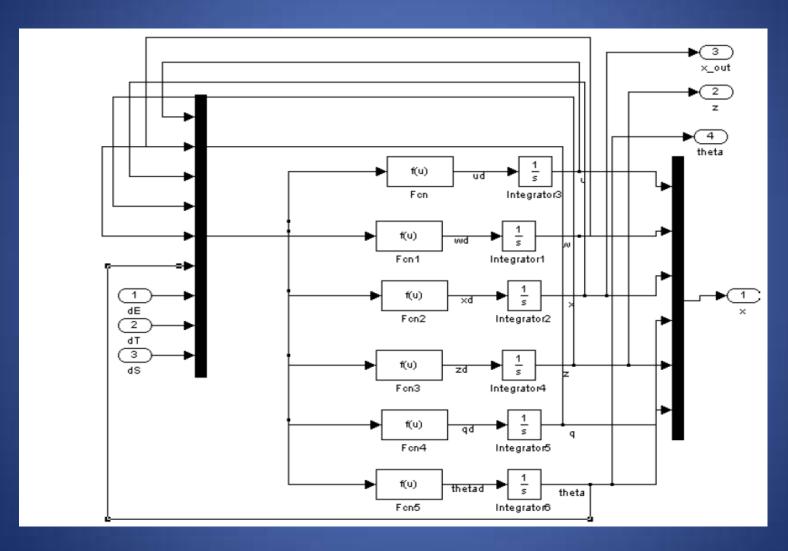
No much difference in results due to ρ that is very small.

- From J1 to J8 (cost is getting higher), and from J1 to J3 (performance is getting better), but from J4 to J8 (no change in performance).
- -J8 is our choice for controlling all the inputs with all state variables of the system with high performance and high cost.
- For future work, we can compromise between J 3and J 4 to have J9 with specific state variables (not all state variables) and also specific inputs for compromising between performance and cost.

LQR Layout:

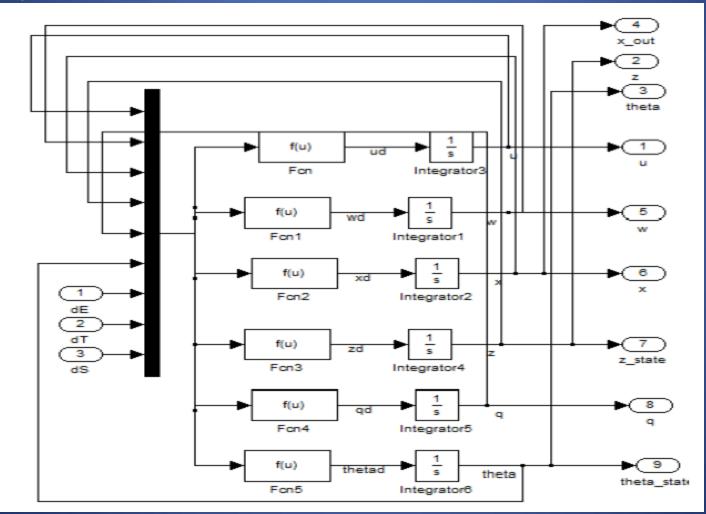


Feedback linearization of the model for LQR layout

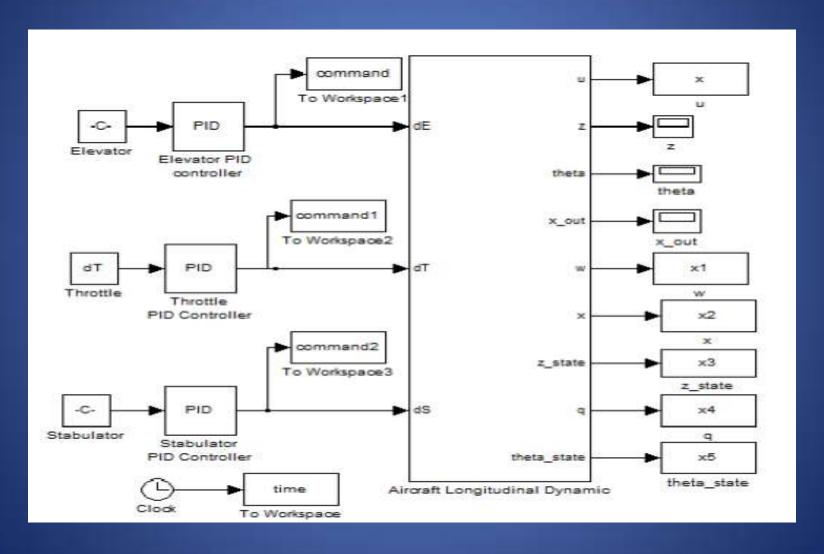


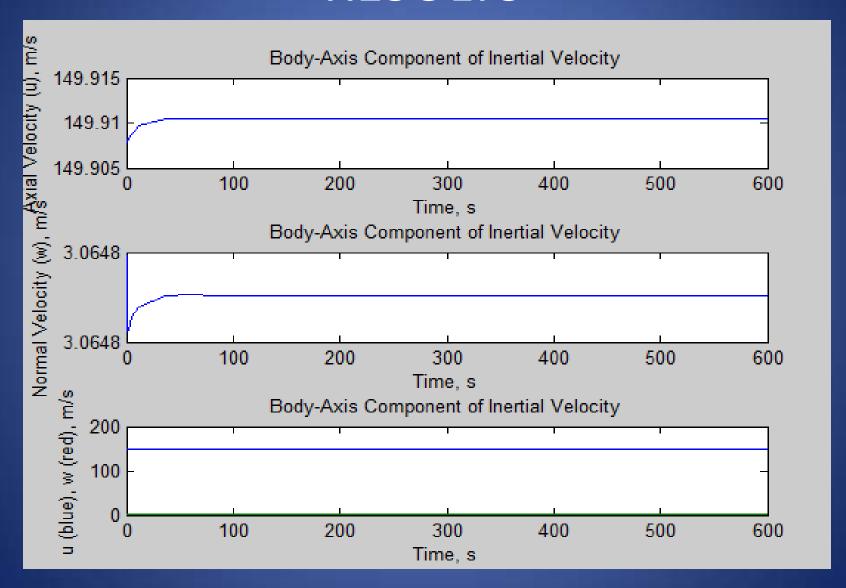
Dynamic model of longitudinal aircraft for LQR layout

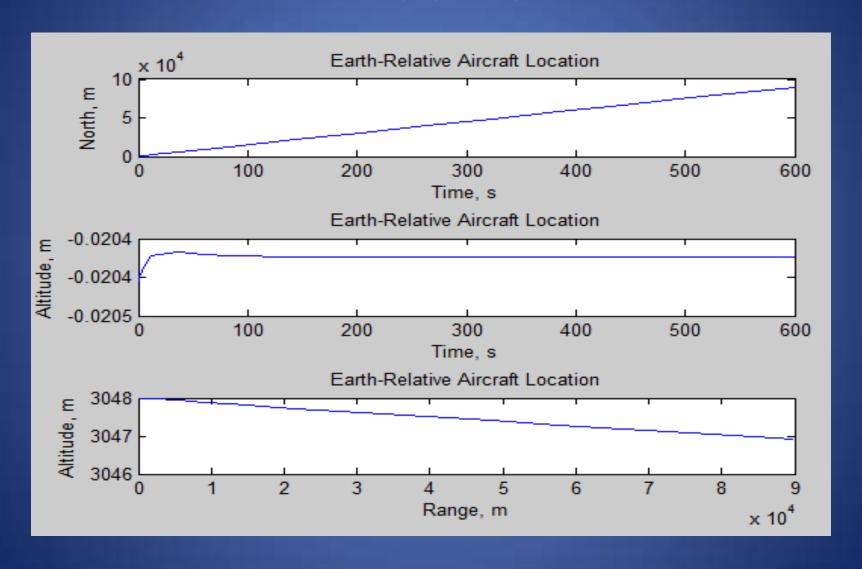
PID Layout:

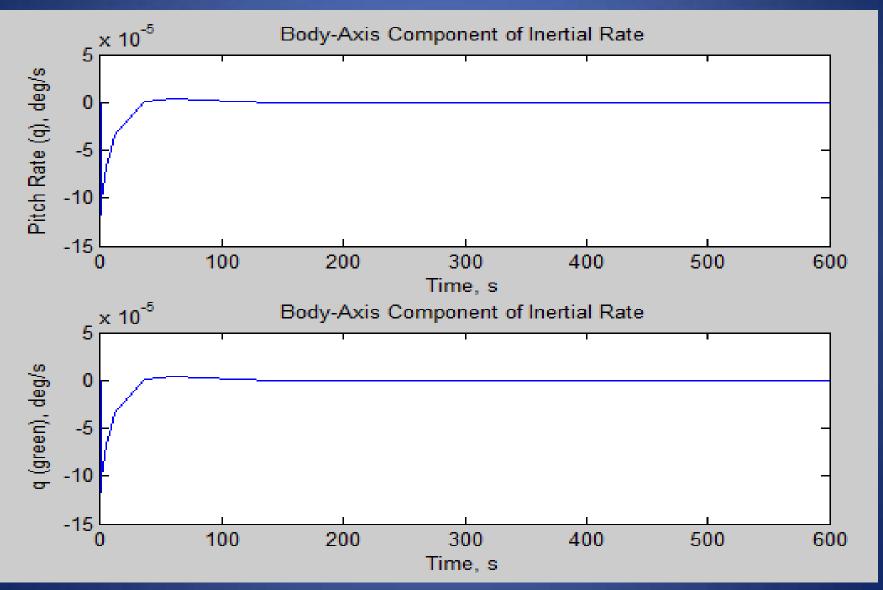


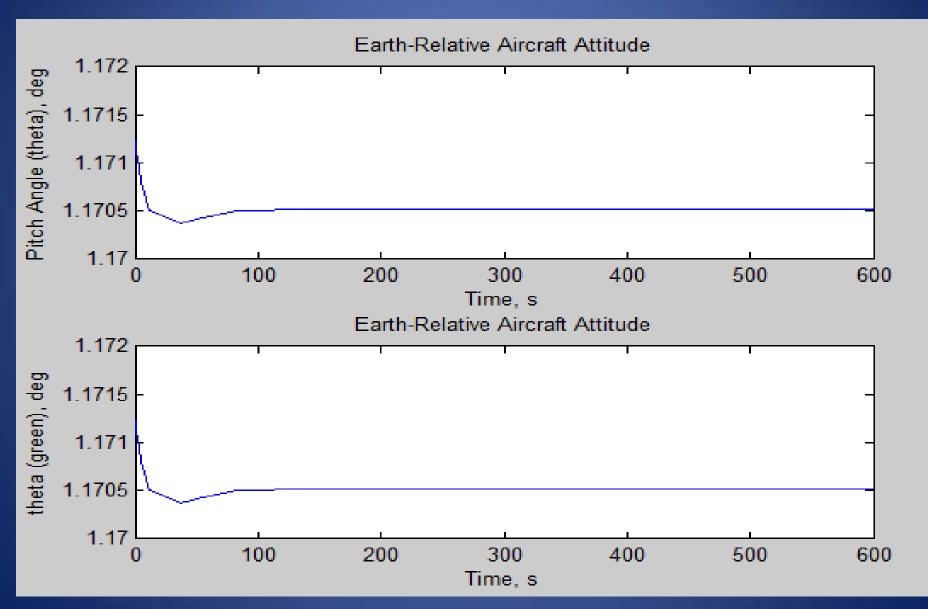
Dynamic model of longitudinal aircraft for PID layout

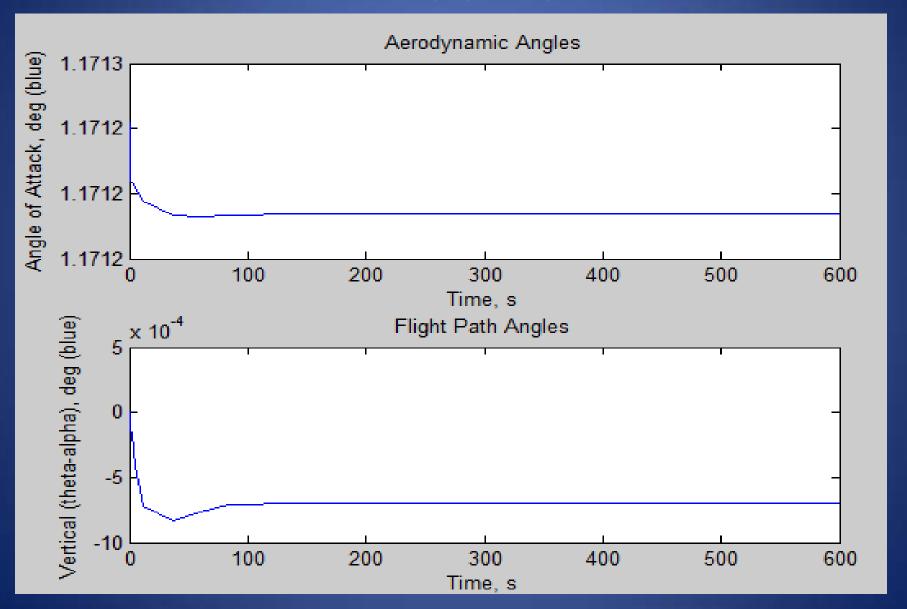




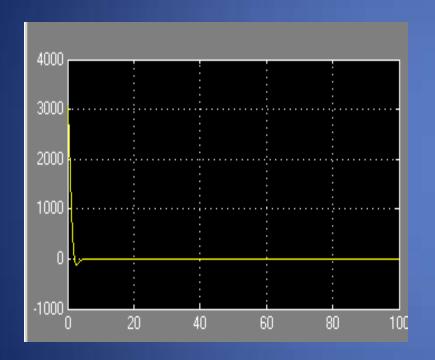




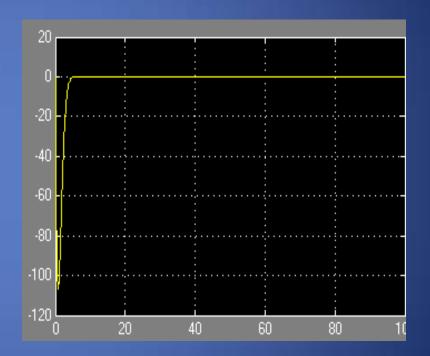




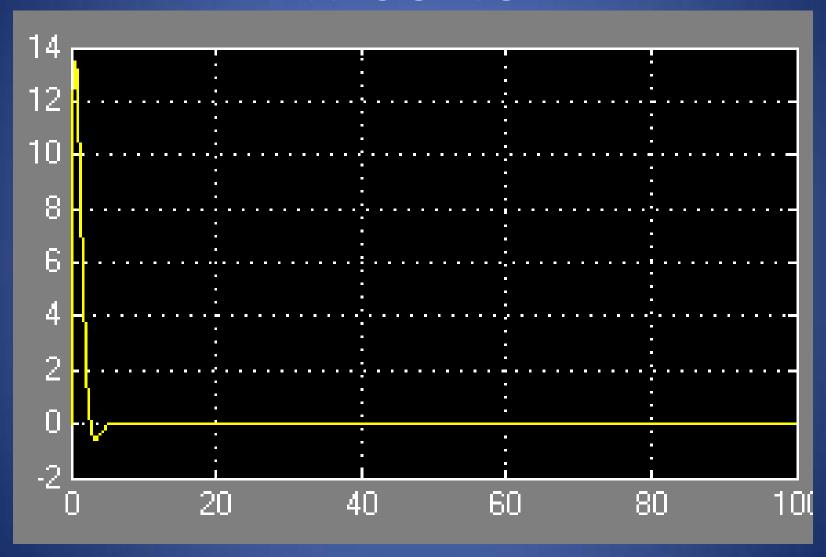
LQR Feedback Results:



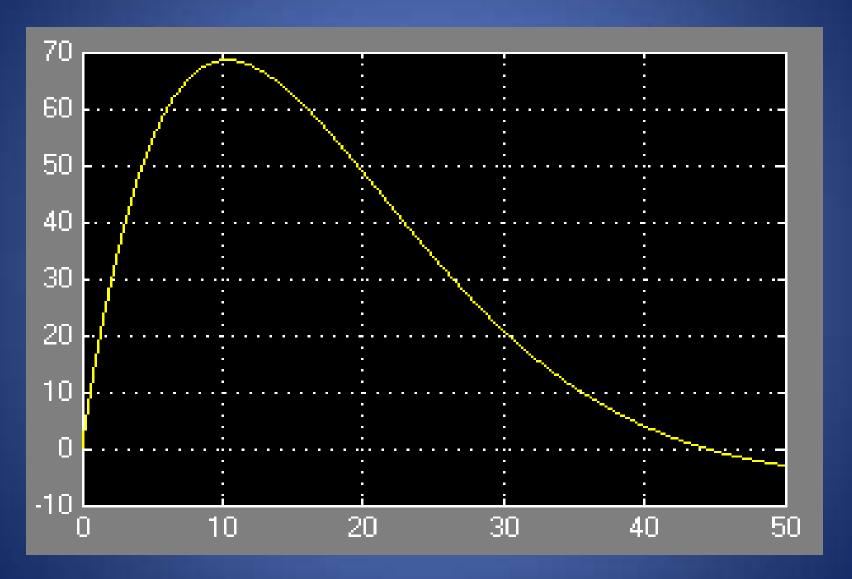
Altitude stabilization



Range stabilization



Pitch angle stabilization



Range stabilization

THANK YOU

Done By: Abbas Chamas

Fadi Al Fara

Bilal Jaroudi

Moustafa Moustafa